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Low Dropout Regulator - CMOS 150 mA

NCP103

The NCP103 is 150 mA LDO that provides the engineer with a very stable, accurate voltage with low noise suitable for space constrained, noise sensitive applications. In order to optimize performance for battery operated portable applications, the NCP103 employs the dynamic quiescent current adjustment for very low I_Q consumption at no−load.

Features

- Operating Input Voltage Range: 1.7 V to 5.5 V
- Available in Fixed Voltage Options: 0.9 V to 3.5 V Contact Factory for Other Voltage Options
- Very Low Quiescent Current of Typ. 50 μA
- Standby Current Consumption: Typ. 0.1 μA
- Low Dropout: 75 mV Typical at 150 mA
- ±1% Accuracy at Room Temperature
- High Power Supply Ripple Rejection: 75 dB at 1 kHz
- Thermal Shutdown and Current Limit Protections
- Stable with a 1 μF Ceramic Output Capacitor
- Available in uDFN 1.0 x 1.0 mm Package
- These Devices are Pb−Free, Halogen Free/BFR Free and are RoHS Compliant

Typical Applicaitons

- PDAs, Mobile phones, GPS, Smartphones
- Wireless Handsets, Wireless LAN, Bluetooth®, Zigbee®
- Portable Medical Equipment
- Other Battery Powered Applications

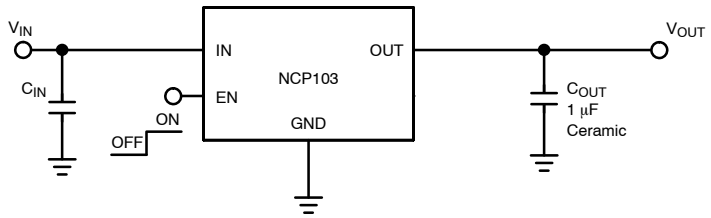
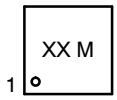


Figure 1. Typical Application Schematic



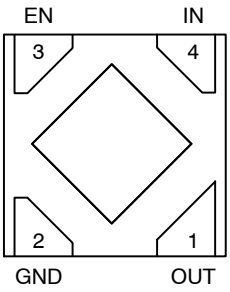
UDFN4
MX SUFFIX
CASE 517CU

MARKING DIAGRAM



XX = Specific Device Code
M = Date Code

PIN CONNECTION



(Bottom View)

ORDERING INFORMATION

See detailed ordering, marking and shipping information on page 13 of this data sheet.

NOTE: Some of the devices on this data sheet have been **DISCONTINUED**. Please refer to the table on page 13.

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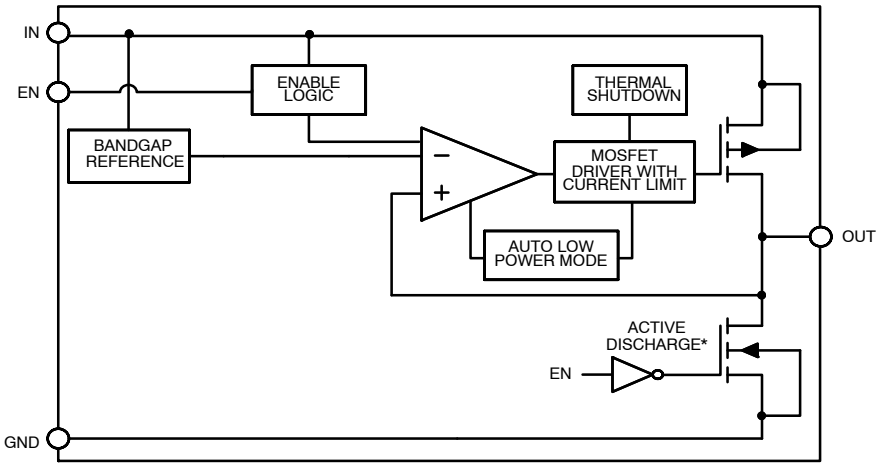

*Active output discharge function is present only in NCP103AMXyyyTCG devices. yyy denotes the particular V_{OUT} option.

Figure 2. Simplified Schematic Block Diagram

PIN FUNCTION DESCRIPTION

Pin No.	Pin Name	Description
1	OUT	Regulated output voltage pin. A small ceramic capacitor with minimum value of 1 μF is needed from this pin to ground to assure stability.
2	GND	Power supply ground.
3	EN	Driving EN over 0.9 V turns on the regulator. Driving EN below 0.4 V puts the regulator into shutdown mode.
4	IN	Input pin. A small capacitor is needed from this pin to ground to assure stability.
–	EPAD	Exposed pad should be connected directly to the GND pin. Soldered to a large ground copper plane allows for effective heat removal.

ABSOLUTE MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Input Voltage (Note 1)	V_{IN}	−0.3 V to 6 V	V
Output Voltage	V_{OUT}	−0.3 V to V_{IN} + 0.3 V or 6 V	V
Enable Input	V_{EN}	−0.3 V to V_{IN} + 0.3 V or 6 V	V
Output Short Circuit Duration	t_{SC}	∞	s
Maximum Junction Temperature	$T_{J(MAX)}$	150	°C
Storage Temperature	T_{STG}	−55 to 150	°C
ESD Capability, Human Body Model (Note 2)	ESD_{HBM}	2000	V
ESD Capability, Machine Model (Note 2)	ESD_{MM}	200	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTIS and APPLICATION INFORMATION for Safe Operating Area.
2. This device series incorporates ESD protection and is tested by the following methods:
   ESD Human Body Model tested per EIA/JESD22−A114,
   ESD Machine Model tested per EIA/JESD22−A115,
   Latchup Current Maximum Rating tested per JEDEC standard: JESD78.

THERMAL CHARACTERISTICS (Note 3)

Rating	Symbol	Value	Unit
Thermal Characteristics, uDFN4 1x1 mm Thermal Resistance, Junction−to−Air	$R_{\theta JA}$	170	°C/W

3. Single component mounted on 1 oz, FR 4 PCB with 645 mm^2 Cu area.

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ELECTRICAL CHARACTERISTICS $-40°C \le T_J \le 85°C$; $V_{IN} = V_{OUT(NOM)} + 1$ V for V_{OUT} options greater than 1.5 V. Otherwise V_{IN} = 2.5 V, whichever is greater; I_{OUT} = 1 mA, $C_{IN} = C_{OUT}$ = 1 μF, unless otherwise noted. V_{EN} = 0.9 V. Typical values are at T_J = +25°C. Min./Max. are for T_J = −40°C and T_J = +85°C respectively.

Parameter	Test Conditions		Symbol	Min	Typ	Max	Unit
Operating Input Voltage			V_{IN}	1.7		5.5	V
Output Voltage Accuracy	$-40°C \le T_J \le 85°C$	$V_{OUT} \le 2.0$ V	V_{OUT}	−40		+40	mV
		$V_{OUT} > 2.0$ V		−2		+2	%
Line Regulation	V_{OUT} + 0.5 V ≤ V_{IN} ≤ 5.5 V ($V_{IN} \ge$ 1.7 V)		Reg_{LINE}		0.01	0.1	%/V
Load Regulation	I_{OUT} = 1 mA to 150 mA		Reg_{LOAD}		10	30	mV
Load Transient	I_{OUT} = 1 mA to 150 mA or 150 mA to 1 mA in 1 μs, C_{OUT} = 1 μF		$Tran_{LOAD}$		−30/ +20		mV
Dropout Voltage (Note 4)	I_{OUT} = 150 mA	V_{OUT} = 1.5 V	V_{DO}		180	235	mV
		V_{OUT} = 1.85 V			120	165	
		V_{OUT} = 2.8 V			75	125	
		V_{OUT} = 3.0 V			72	120	
		V_{OUT} = 3.1 V			70	120	
		V_{OUT} = 3.3 V			65	110	
Output Current Limit	V_{OUT} = 90% $V_{OUT(nom)}$		I_{CL}	150	550		mA
Ground Current	I_{OUT} = 0 mA		I_Q		50	95	μA
Shutdown Current	$V_{EN} \le 0.4$ V, V_{IN} = 5.5 V		I_{DIS}		0.01	1	μA
EN Pin Threshold Voltage High Threshold Low Threshold	 V_{EN} Voltage increasing V_{EN} Voltage decreasing		 V_{EN_HI} V_{EN_LO}	 0.9		 0.4	V
EN Pin Input Current	V_{EN} = 5.5 V		I_{EN}		0.3	1.0	μA
Power Supply Rejection Ratio	V_{IN} = 3.6 V, V_{OUT} = 3.1 V I_{OUT} = 150 mA	f = 1 kHz	PSRR		75		dB
Output Noise Voltage	V_{IN} = 2.5 V, V_{OUT} = 1.8 V, I_{OUT} = 150 mA f = 10 Hz to 100 kHz		V_N		60		μV_{rms}
Thermal Shutdown Temperature	Temperature increasing from T_J = +25°C		T_{SD}		160		°C
Thermal Shutdown Hysteresis	Temperature falling from T_{SD}		T_{SDH}		20		°C
Active Output Discharge Resistance	V_{EN} < 0.4 V, Version A only		R_{DIS}		100		Ω

4. Characterized when V_{OUT} falls 100 mV below the regulated voltage at $V_{IN} = V_{OUT(NOM)} + 1$ V.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

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TYPICAL CHARACTERISTICS

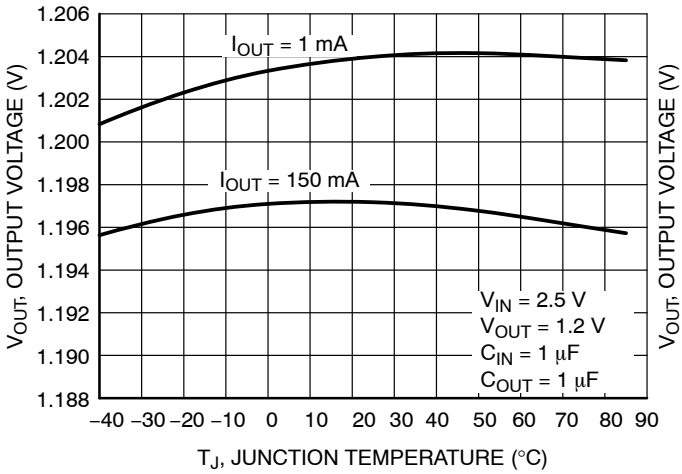

Figure 3. Output Voltage vs. Temperature
V_{OUT} = 1.2 V

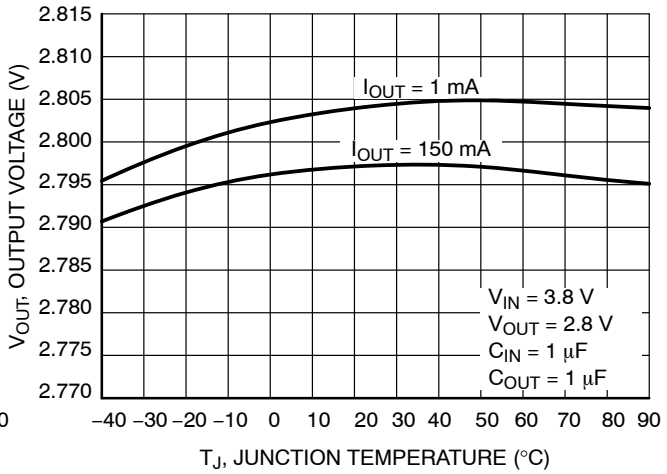

Figure 4. Output Voltage vs. Temperature
V_{OUT} = 2.8 V

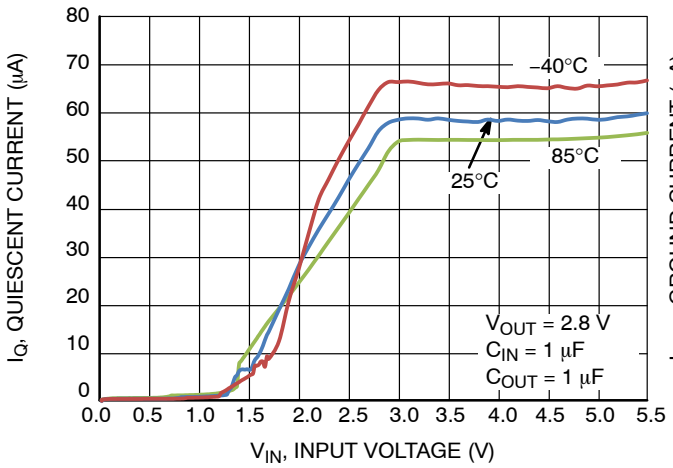

Figure 5. Quiescent Current vs. Input Voltage

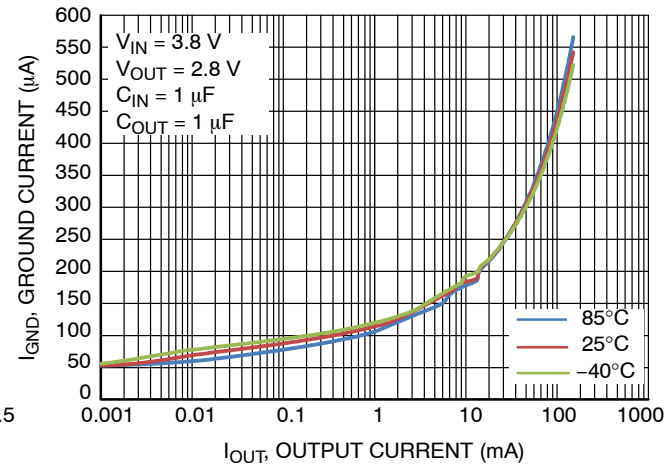

Figure 6. Ground Current vs. Output Current

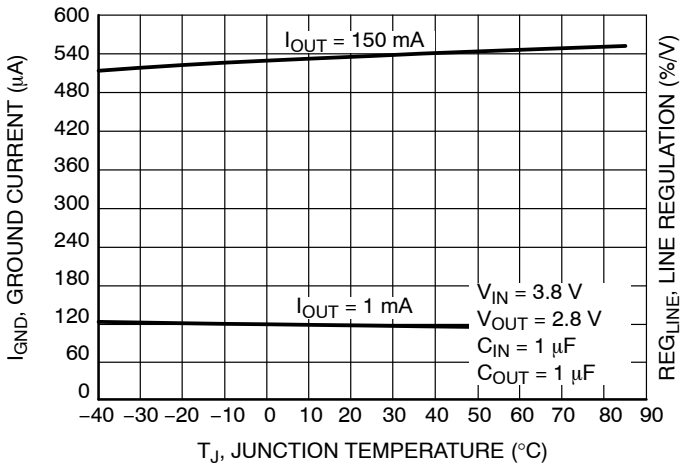

Figure 7. Ground Current vs. Temperature

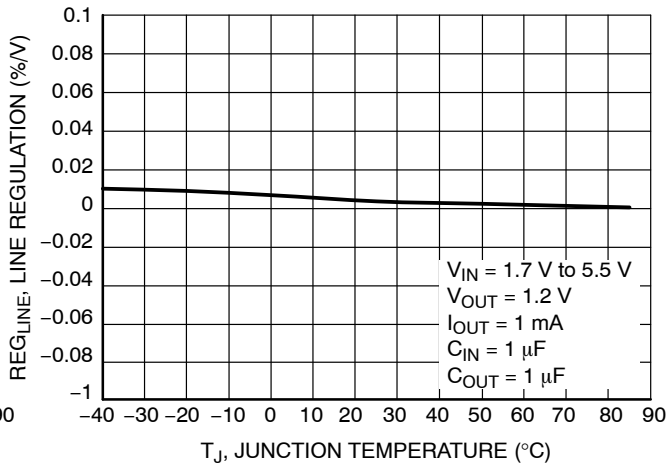

Figure 8. Line Regulation vs. Output Current
V_{OUT} = 1.2 V

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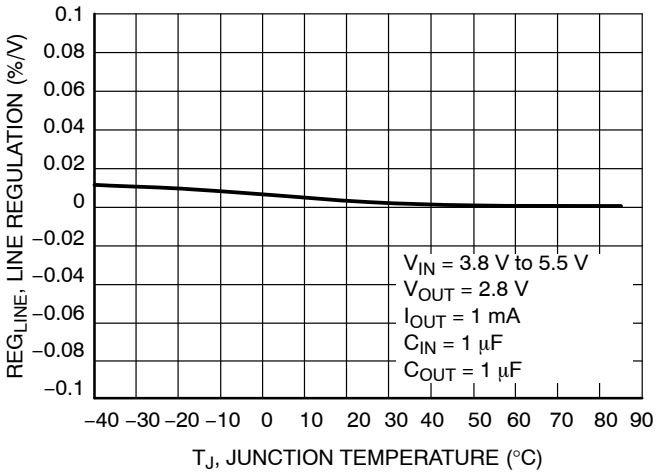

Figure 9. Line Regulation vs. Temperature
V_{OUT} = 2.8 V

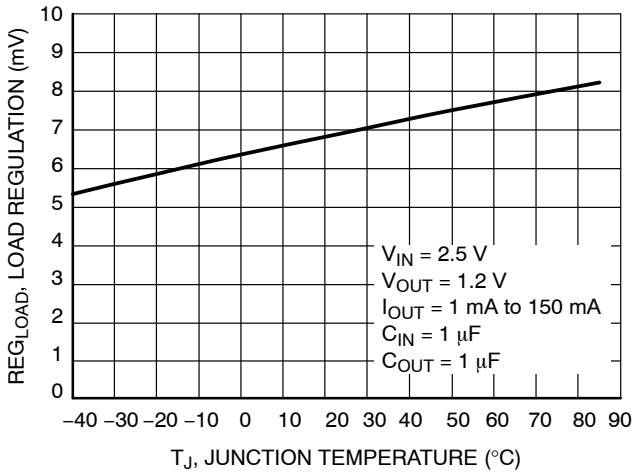

Figure 10. Load Regulation vs. Temperature
V_{OUT} = 1.2 V

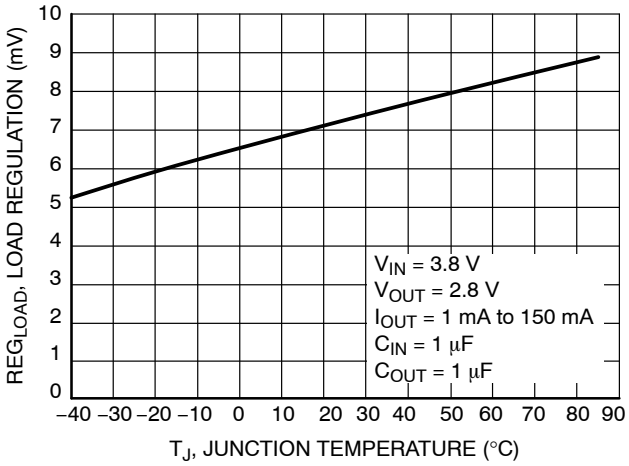

Figure 11. Load Regulation vs. Temperature
V_{OUT} = 2.8 V

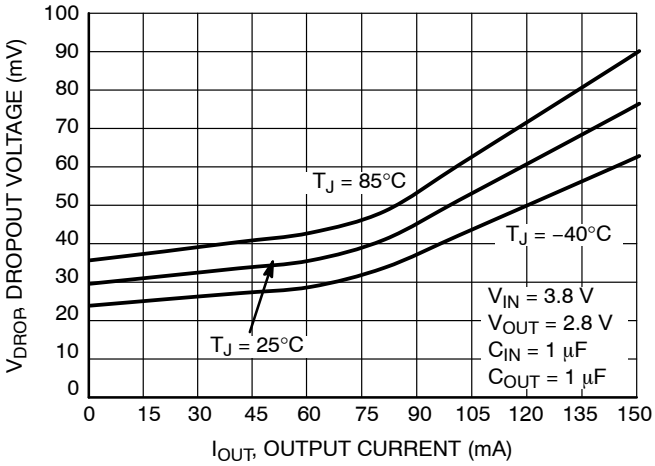

Figure 12. Dropout Voltage vs. Output Current
V_{OUT} = 2.8 V

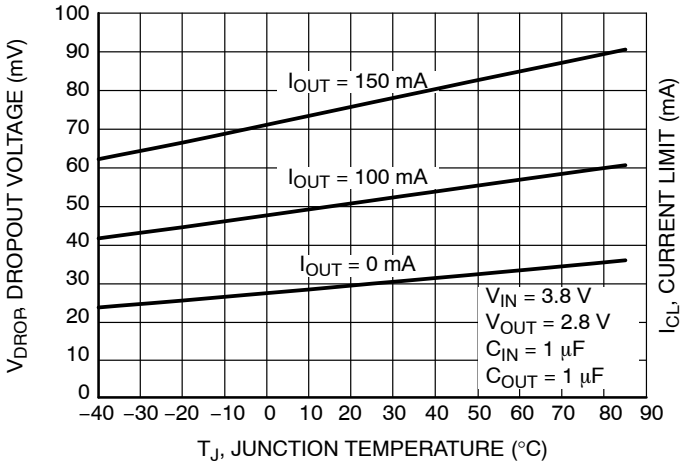

Figure 13. Dropout Voltage vs. Temperature

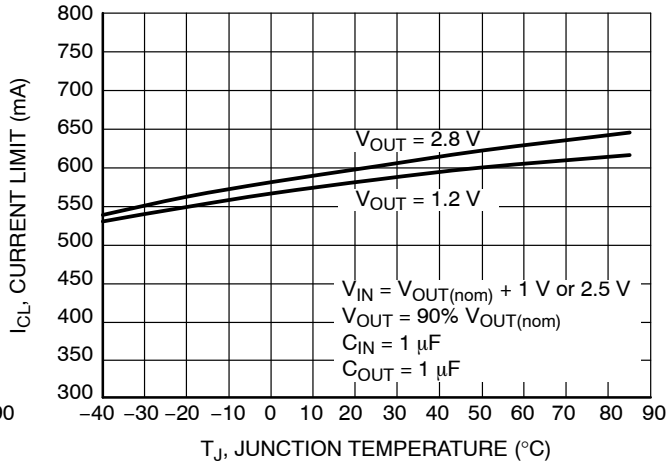

Figure 14. Current Limit vs. Temperature

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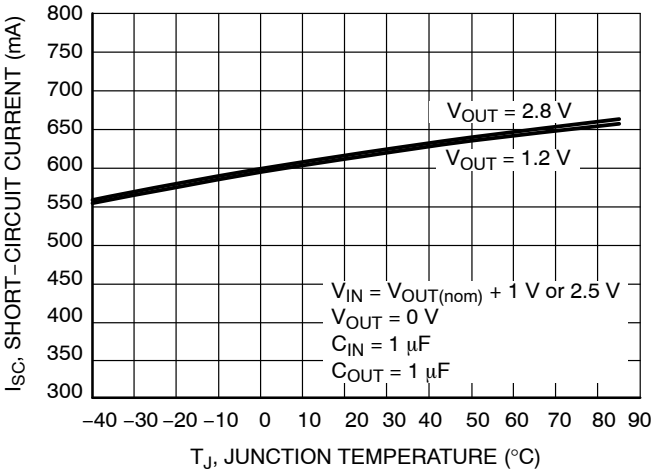

Figure 15. Short−Circuit Current vs. Temperature

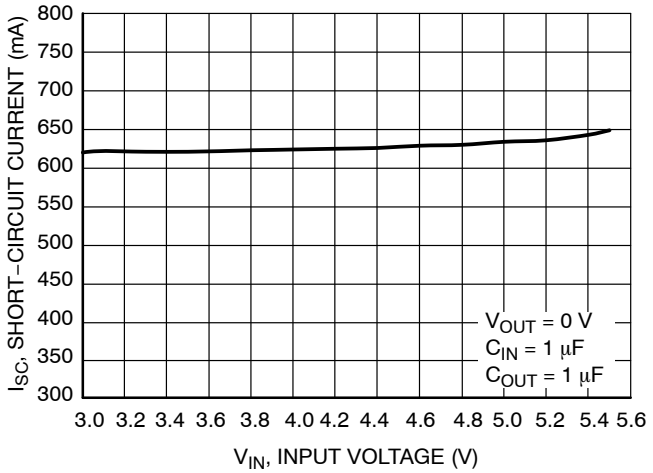

Figure 16. Short−Circuit Current vs. Input Voltage

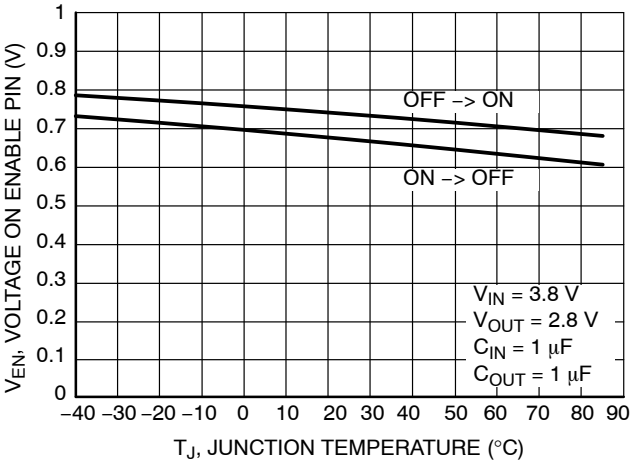

Figure 17. Enable Voltage Threshold vs. Temperature

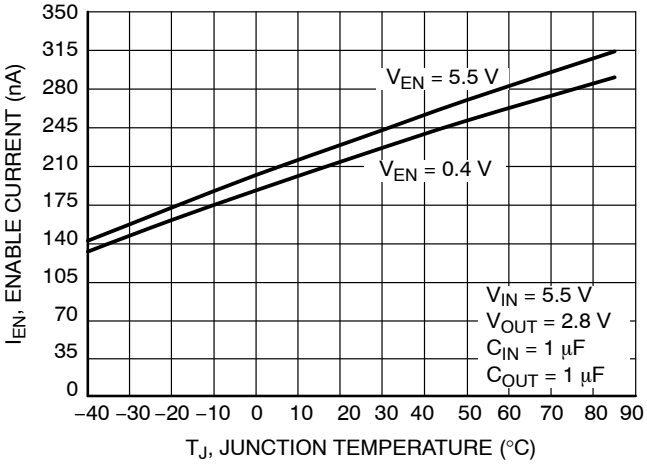

Figure 18. Current to Enable Pin vs. Temperature

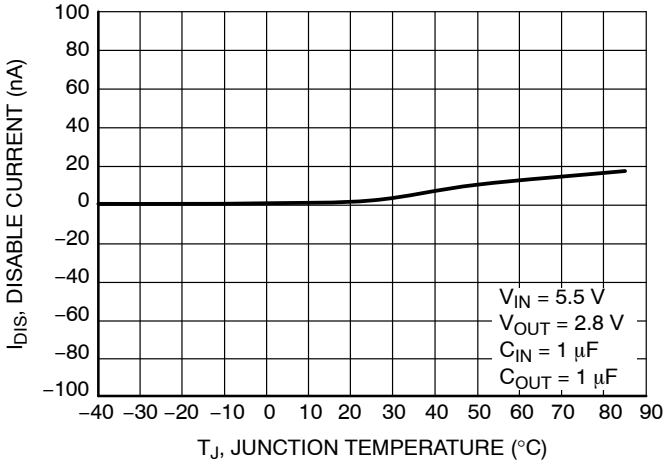

Figure 19. Disable Current vs. Temperature

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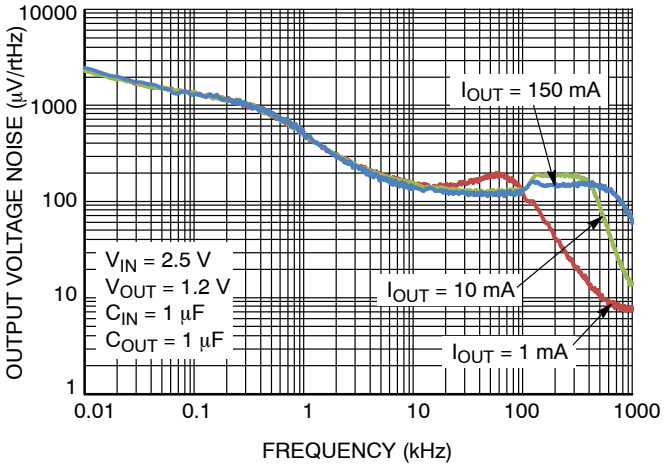
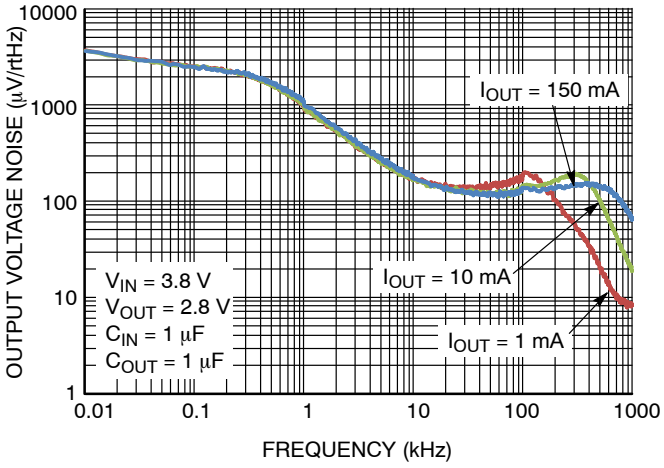


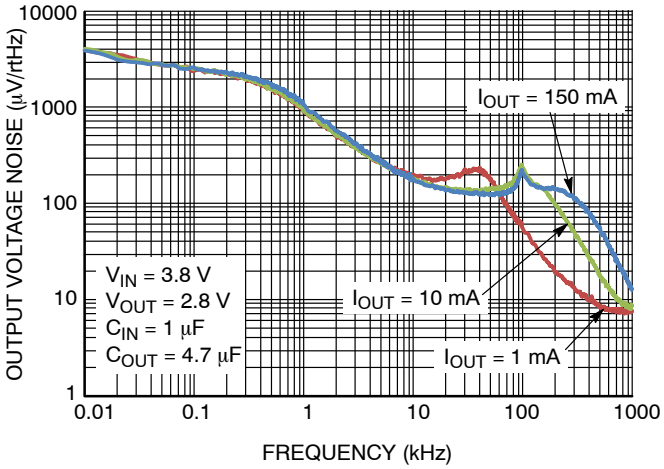
I_{OUT}	RMS Output Noise (μV)	
	10 Hz – 100 kHz	100 Hz – 100 kHz
1 mA	60.93	59.11
10 mA	52.73	50.63
150 mA	51.20	48.96

Figure 20. Output Voltage Noise Spectral Density for V_{OUT} = 1.2 V, C_{OUT} = 1 μF



I_{OUT}	RMS Output Noise (μV)	
	10 Hz – 100 kHz	100 Hz – 100 kHz
1 mA	79.23	74.66
10 mA	75.03	70.37
150 mA	77.28	72.66

Figure 21. Output Voltage Noise Spectral Density for V_{OUT} = 2.8 V, C_{OUT} = 1 μF



I_{OUT}	RMS Output Noise (μV)	
	10 Hz – 100 kHz	100 Hz – 100 kHz
1 mA	80.17	75.29
10 mA	81.28	76.46
150 mA	81.31	76.77

Figure 22. Output Voltage Noise Spectral Density for V_{OUT} = 2.8 V, C_{OUT} = 4.7 μF

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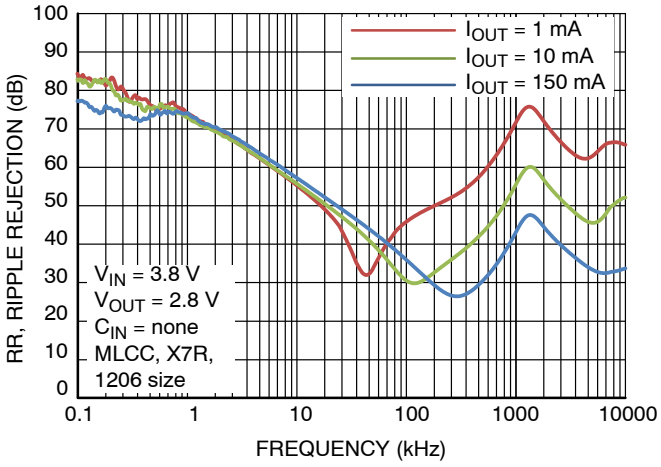

Figure 23. Power Supply Rejection Ratio, V_{OUT} = 1.2 V, C_{OUT} = 1 μF

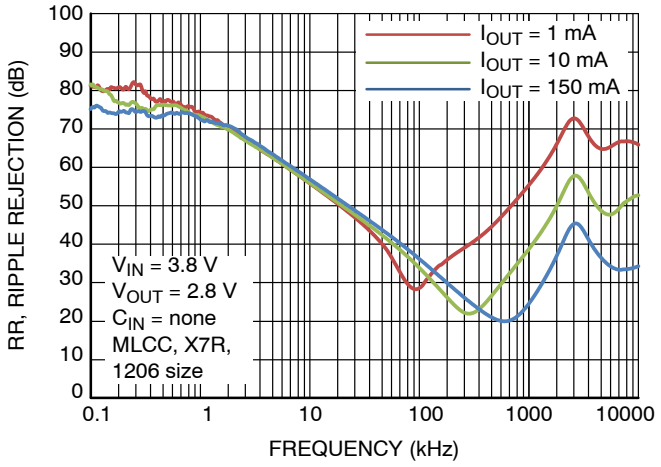

Figure 24. Power Supply Rejection Ratio, V_{OUT} = 2.8 V, C_{OUT} = 4.7 μF

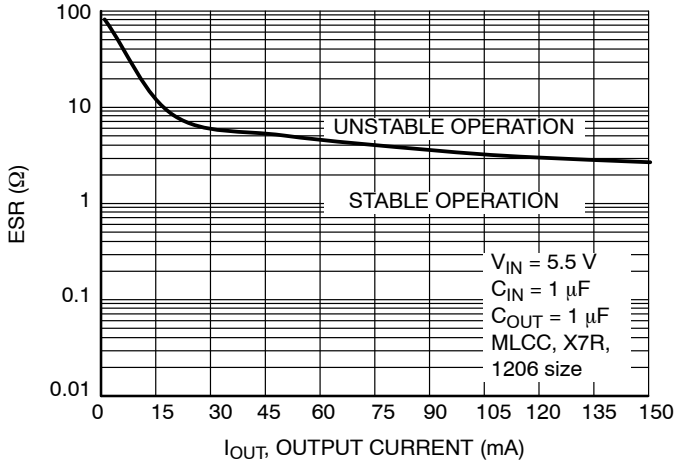

Figure 25. Output Capacitor ESR vs. Output Current

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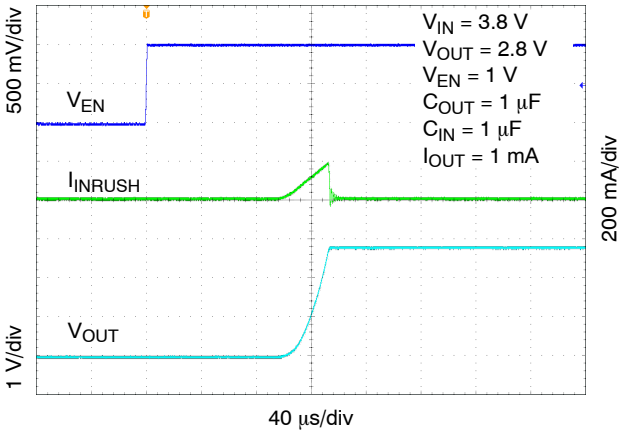

Figure 26. Enable Turn−on Response, C_{OUT} = 1 μF, I_{OUT} = 1 mA

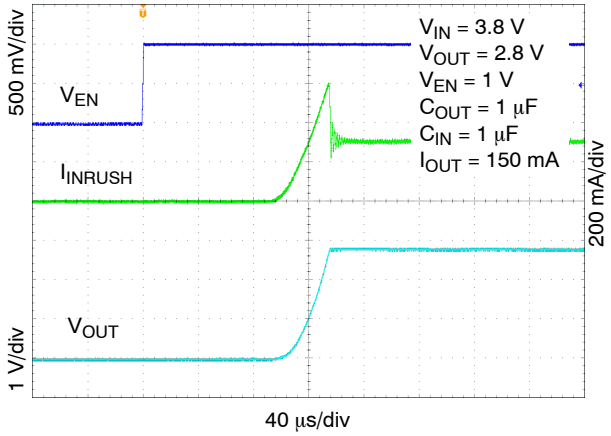

Figure 27. Enable Turn−on Response, C_{OUT} = 1 μF, I_{OUT} = 150 mA

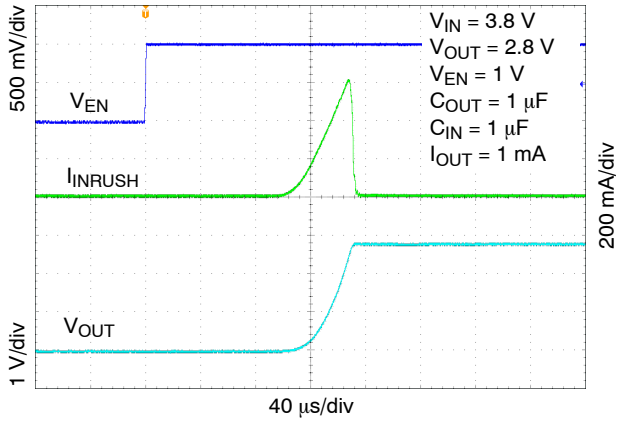

Figure 28. Enable Turn−on Response, C_{OUT} = 4.7 μF, I_{OUT} = 1 mA

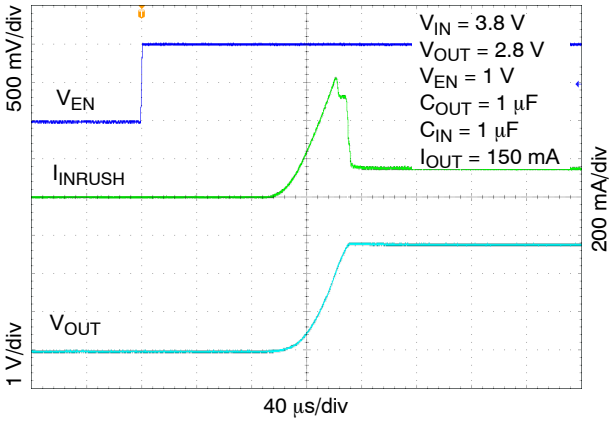

Figure 29. Enable Turn−on Response, C_{OUT} = 4.7 μF, I_{OUT} = 150 mA

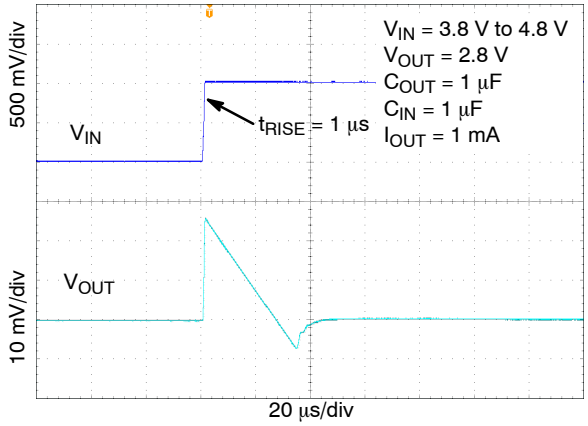

Figure 30. Line Transient Response − Rising Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA

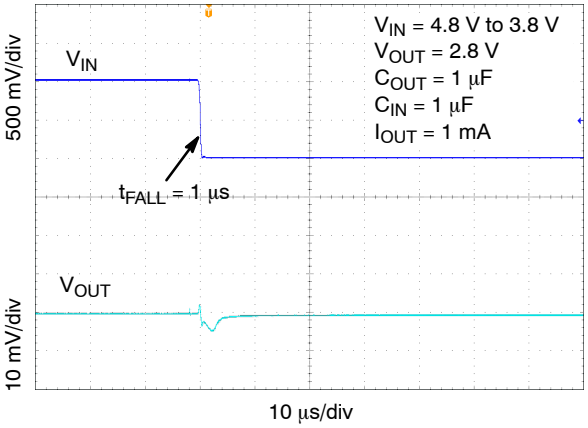

Figure 31. Line Transient Response − Falling Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA

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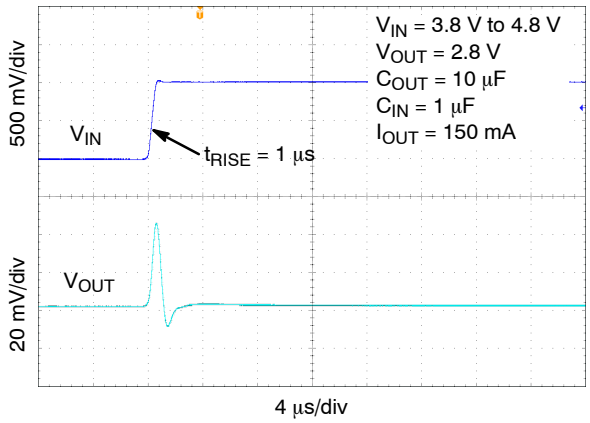

Figure 32. Line Transient Response – Rising Edge, V_{OUT} = 2.8 V, I_{OUT} = 150 mA

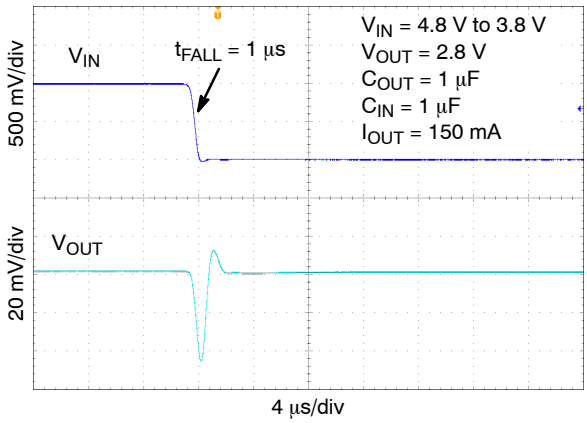

Figure 33. Line Transient Response – Falling Edge, V_{OUT} = 2.8 V, I_{OUT} = 150 mA

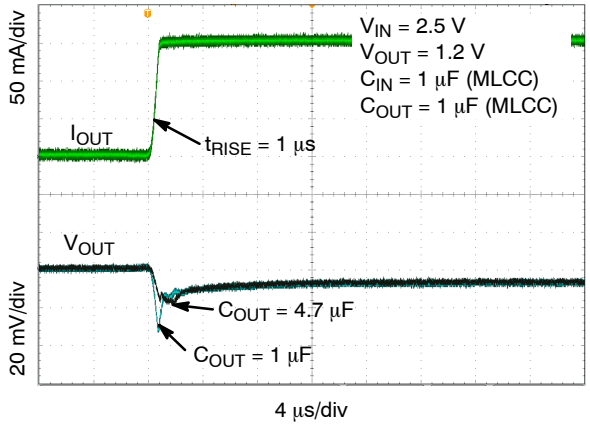

Figure 34. Load Transient Response – Rising Edge, V_{OUT} = 1.2 V, I_{OUT} = 1 mA to 150 mA, C_{OUT} = 1 μF, 4.7 μF

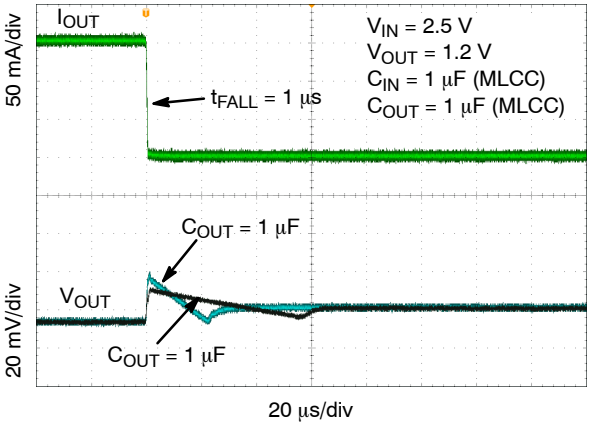

Figure 35. Load Transient Response – Falling Edge, V_{OUT} = 1.2 V, I_{OUT} = 1 mA to 150 mA, C_{OUT} = 1 μF, 4.7 μF

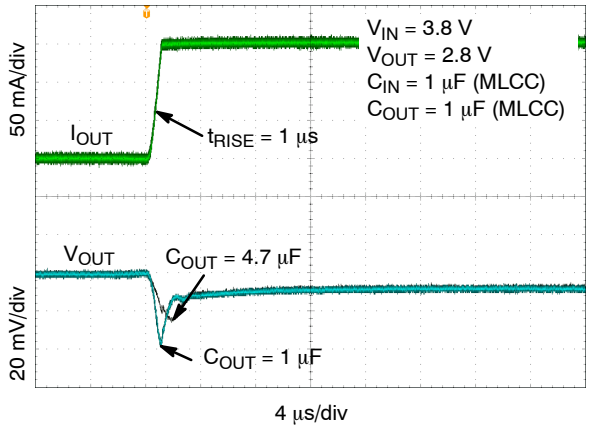

Figure 36. Load Transient Response – Rising Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA to 150 mA, C_{OUT} = 1 μF, 4.7 μF

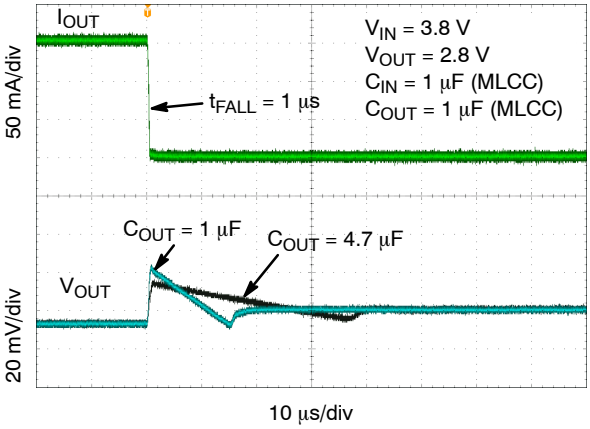

Figure 37. Load Transient Response – Falling Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA to 150 mA, C_{OUT} = 1 μF, 4.7 μF

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TYPICAL CHARACTERISTICS

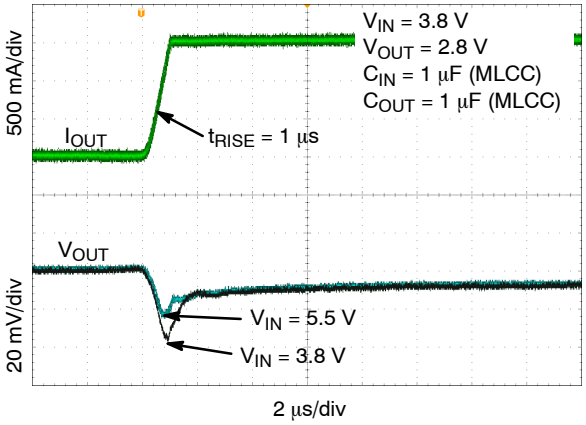

Figure 38. Load Transient Response – Rising Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA to 150 mA, V_{IN} = 3.8 V, 5.5 V

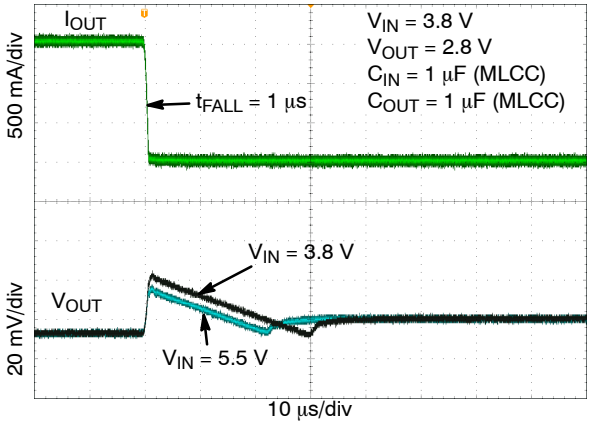

Figure 39. Load Transient Response – Falling Edge, V_{OUT} = 2.8 V, I_{OUT} = 1 mA to 150 mA, V_{IN} = 3.8 V, 5.5 V

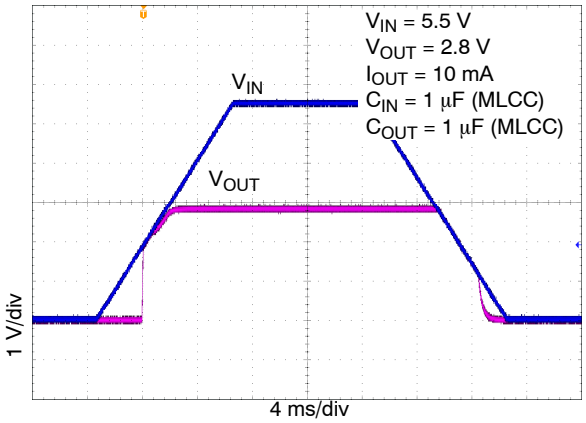

Figure 40. Turn−on/off – Slow Rising V_{IN}

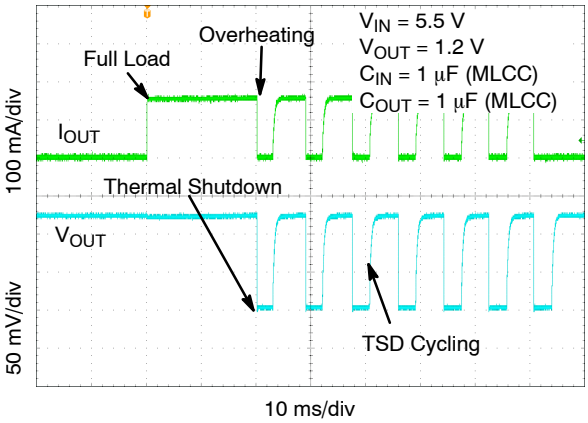

Figure 41. Short−Circuit and Thermal Shutdown

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APPLICATIONS INFORMATION

General

The NCP103 is a high performance 150 mA Low Dropout Linear Regulator. This device delivers very high PSRR (over 75 dB at 1 kHz) and excellent dynamic performance as load/line transients. In connection with very low quiescent current this device is very suitable for various battery powered applications such as tablets, cellular phones, wireless and many others. The device is fully protected in case of output overload, output short circuit condition and overheating, assuring a very robust design.

Input Capacitor Selection (C_{IN})

It is recommended to connect at least a 1μF Ceramic X5R or X7R capacitor as close as possible to the IN pin of the device. This capacitor will provide a low impedance path for unwanted AC signals or noise modulated onto constant input voltage. There is no requirement for the min. /max. ESR of the input capacitor but it is recommended to use ceramic capacitors for their low ESR and ESL. A good input capacitor will limit the influence of input trace inductance and source resistance during sudden load current changes. Larger input capacitor may be necessary if fast and large load transients are encountered in the application.

Output Decoupling (C_{OUT})

The NCP103 requires an output capacitor connected as close as possible to the output pin of the regulator. The recommended capacitor value is 1 μF and X7R or X5R dielectric due to its low capacitance variations over the specified temperature range. The NCP103 is designed to remain stable with minimum effective capacitance of 0.22 μF to account for changes with temperature, DC bias and package size. Especially for small package size capacitors such as 0402 the effective capacitance drops rapidly with the applied DC bias.

There is no requirement for the minimum value of Equivalent Series Resistance (ESR) for the C_{OUT} but the maximum value of ESR should be less than 3 Ω. Larger output capacitors and lower ESR could improve the load transient response or high frequency PSRR. It is not recommended to use tantalum capacitors on the output due to their large ESR. The equivalent series resistance of tantalum capacitors is also strongly dependent on the temperature, increasing at low temperature.

Enable Operation

The NCP103 uses the EN pin to enable/disable its device and to deactivate/activate the active discharge function.

If the EN pin voltage is <0.4 V the device is guaranteed to be disabled. The pass transistor is turned−off so that there is virtually no current flow between the IN and OUT. The active discharge transistor is active so that the output voltage V_{OUT} is pulled to GND through a 100 Ω resistor. In the disable state the device consumes as low as typ. 10 nA from the V_{IN}.

If the EN pin voltage >0.9 V the device is guaranteed to be enabled. The NCP103 regulates the output voltage and the active discharge transistor is turned−off.

The EN pin has internal pull−down current source with typ. value of 300 nA which assures that the device is turned−off when the EN pin is not connected. In the case where the EN function isn't required the EN should be tied directly to IN.

Output Current Limit

Output Current is internally limited within the IC to a typical 550 mA. The NCP103 will source this amount of current measured with a voltage drops on the 90% of the nominal V_{OUT}. If the Output Voltage is directly shorted to ground (V_{OUT} = 0 V), the short circuit protection will limit the output current to 580 mA (typ). The current limit and short circuit protection will work properly over whole temperature range and also input voltage range. There is no limitation for the short circuit duration.

Thermal Shutdown

When the die temperature exceeds the Thermal Shutdown threshold (T_{SD} − 160°C typical), Thermal Shutdown event is detected and the device is disabled. The IC will remain in this state until the die temperature decreases below the Thermal Shutdown Reset threshold (T_{SDU} − 140°C typical). Once the IC temperature falls below the 140°C the LDO is enabled again. The thermal shutdown feature provides the protection from a catastrophic device failure due to accidental overheating. This protection is not intended to be used as a substitute for proper heat sinking.

Power Dissipation

As power dissipated in the NCP103 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature rise for the part.

The maximum power dissipation the NCP103 can handle is given by:

$$P_{D(MAX)} = \frac{\left[125^{\circ}C - T_A\right]}{\theta_{JA}} \qquad \text{(eq. 1)}$$

The power dissipated by the NCP103 for given application conditions can be calculated from the following equations:

$$P_D \approx V_{IN}\left(I_{GND}@I_{OUT}\right) + I_{OUT}\left(V_{IN} - V_{OUT}\right) \qquad \text{(eq. 2)}$$

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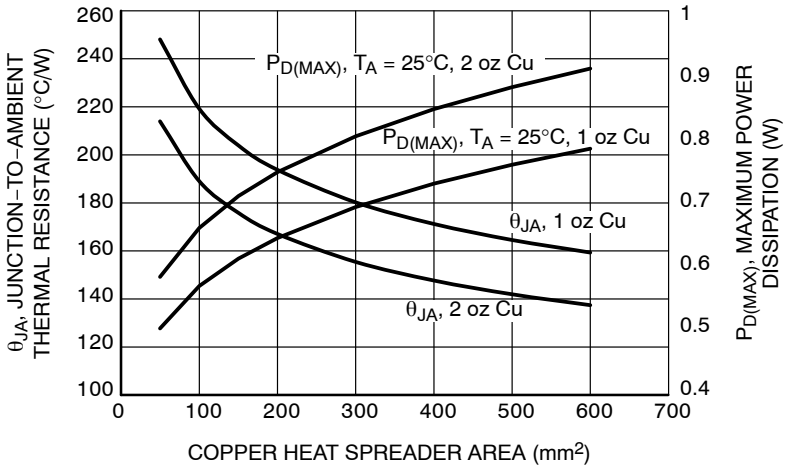

Figure 42. θ_{JA} vs. Copper Area (uDFN4)

Reverse Current

The PMOS pass transistor has an inherent body diode which will be forward biased in the case that $V_{OUT} > V_{IN}$. Due to this fact in cases, where the extended reverse current condition can be anticipated the device may require additional external protection.

Power Supply Rejection Ratio

The NCP103 features very good Power Supply Rejection ratio. If desired the PSRR at higher frequencies in the range 100 kHz − 10 MHz can be tuned by the selection of C_{OUT} capacitor and proper PCB layout.

Turn−On Time

The turn−on time is defined as the time period from EN assertion to the point in which V_{OUT} will reach 98% of its nominal value. This time is dependent on various application conditions such as $V_{OUT(NOM)}$, C_{OUT} and T_A. For example typical value for V_{OUT} = 1.2 V, C_{OUT} = 1 μF, I_{OUT} = 1 mA and T_A = 25°C is 90 μs.

PCB Layout Recommendations

To obtain good transient performance and good regulation characteristics place C_{IN} and C_{OUT} capacitors close to the device pins and make the PCB traces wide. In order to minimize the solution size, use 0402 capacitors. Larger copper area connected to the pins will also improve the device thermal resistance. The actual power dissipation can be calculated from the equation above (Equation 2). Expose pad should be tied the shortest path to the GND pin.

ORDERING INFORMATION

Device	Voltage Option	Marking	Marking Rotation	Option	Package	Shipping†
NCP103BMX185TCG	1.85 V	CJ	0°	Without active output discharge function	uDFN4 (Pb-Free)	3000 / Tape & Reel
NCP103BMX330TCG	3.3 V	Q	90°			

DISCONTINUED (Note 5)

NCP103AMX180TCG	1.8 V	K	180°	With active output discharge function	uDFN4 (Pb-Free)	3000 / Tape & Reel
NCP103AMX310TCG	3.1 V	P	0°			
NCP103BMX185TBG	1.85 V	CJ	0°	Without active output discharge function		

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

5. **DISCONTINUED:** These devices are not recommended for new design. Please contact your **onsemi** representative for information. The most current information on these devices may be available on www.onsemi.com.

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UDFN4 1.0x1.0, 0.65P
CASE 517CU
ISSUE A

DATE 18 DEC 2014

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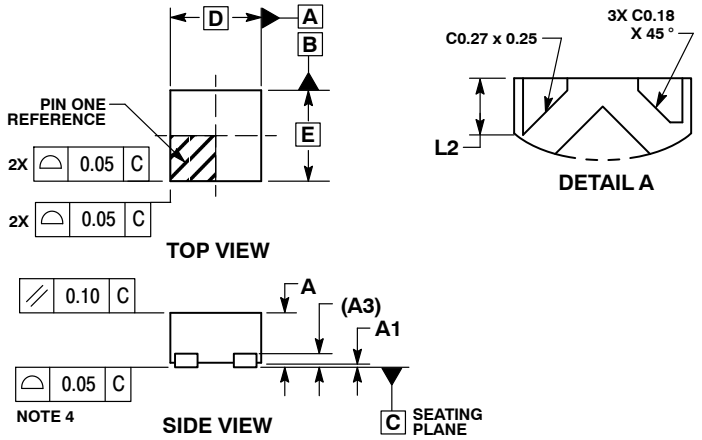
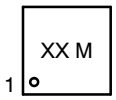


NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. DIMENSION b APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.03 AND 0.07 FROM THE TERMINAL TIPS.
4. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.

	MILLIMETERS	
DIM	MIN	MAX
A	---	0.60
A1	0.00	0.05
A3	0.15 REF	
b	0.20	0.30
D	1.00 BSC	
D2	0.38	0.58
E	1.00 BSC	
e	0.65 BSC	
L	0.20	0.30
L2	0.27	0.37

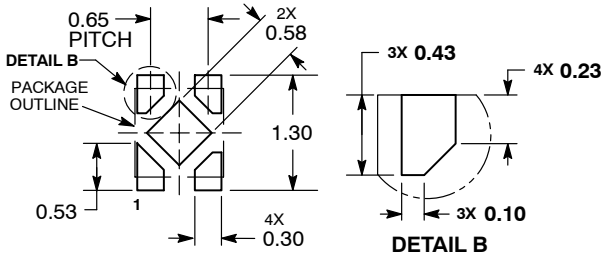
GENERIC MARKING DIAGRAM*



XX = Specific Device Code
M = Date Code

*This information is generic. Please refer to device data sheet for actual part marking.
Pb−Free indicator, “G” or microdot “ ▪”, may or may not be present.

RECOMMENDED MOUNTING FOOTPRINT*



DIMENSIONS: MILLIMETERS

*For additional information on our Pb−Free strategy and soldering details, please download the **onsemi** Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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