TPS763-Q1 Automotive 150mA, 10V, Low-Dropout Linear Regulator

1 Features

- AEC-Q100 Qualified for automotive applications:
  - Temperature Grade 1: –40°C to +150°C T_J (new chip)
  - Device HBM ESD Classification: Level 1C
  - Device CDM ESD Classification: Level C3 (legacy chip)
  - Device CDM ESD Classification: Level C4 (new chip)
- Input voltage range V_{IN}: 2.7V to 10V
- Output voltage range V_{OUT}:
  - Fixed device: 1.6V to 5.0V
- Output voltage range V_{OUT}:
  - Adjustable device: 1.6V to 6.5V
- Output current: up-to 150mA
- Output voltage accuracy:
  - New chip: ±1.5% (typical)
  - Legacy chip: ±2.5% (typical)
- Low quiescent current I_Q:
  - New chip: 65µA (typical) at I_{OUT} = 0mA
  - New chip: 765µA (typical) at I_{OUT} = 150mA
  - Legacy chip: 85µA (typical) at I_{OUT} = 1mA to 150mA
- Dropout voltage:
  - New chip: 175mV (typical) at I_{OUT} = 150mA
  - Legacy chip: 300mV (typical) at I_{OUT} = 150mA
- Supported ESR range:
  - New chip: 0Ω to 1Ω
  - Legacy chip: 0.3Ω to 10Ω
- Thermal shutdown and overcurrent limitation
- Active over-shoot pulldown protection (new chip)
- Operating junction temperature: –40°C to 150°C (new chip)
- 5-pin SOT-23 (DBV):
  - New chip, $R_{\theta JA}$ = 178.6°C/W
  - Legacy chip, $R_{\theta JA}$ = 205.2°C/W

2 Applications

- High-voltage battery system
- Vehicle instrument cluster
- 2-wheeler and 3-wheeler traction drive
- Traction inverter

3 Description

The TPS763xx-Q1 family of low-dropout (LDO) linear voltage regulators supports wide input voltage range of 2.7V to 10V and up-to 150mA of load current. The output range is from 1.6V to 5.0V for fixed version, and 1.6V to 6.5V for adjustable version.

The TPS763xx-Q1 has a ±1.5% output accuracy that is required for powering digital loads with tight supply requirements. The internal soft-start circuit reduces inrush current during start-up (for new chip), thus allowing for smaller input capacitance.

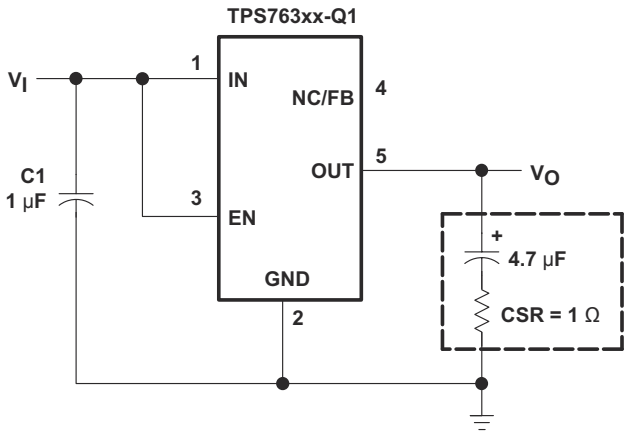
The TPS763xx-Q1 family of low-dropout (LDO) voltage regulators offers the benefits of low-dropout voltage, low-power operation, and miniaturized packaging. These regulators feature low dropout voltages and quiescent currents compared to conventional LDO regulators. Offered in a 5-pin, small outline integrated-circuit SOT-23 package, the TPS763xx-Q1 series devices are and excellent choice for cost-sensitive designs and applications where board space is at a premium.

The TPS763xx-Q1 also features a logic-enabled sleep mode to shut down the regulator, reducing quiescent current to 1µA maximum at T_J = 25°C.

Package Information

PART NUMBER	PACKAGE(1)	PACKAGE SIZE (2)
TPS763xx-Q1	SOT-23 (5)	2.9mm × 2.8mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.
(2) The package size (length × width) is a nominal value and includes pins, where applicable.



Typical Application Circuit

An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.

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4 Pin Configuration and Functions

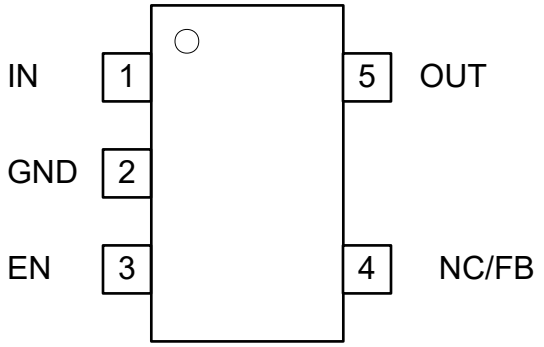


Figure 4-1. DBV Package 5-Pin SOT-23 Top View

Table 4-1. Pin Functions

PIN		TYPE(1)	DESCRIPTION
NAME	NO.		
EN	3	—	Enable pin for the LDO. Driving the EN pin high enables the device. Driving this pin low disables the device. High and low thresholds are listed in the Section 5.5 table. Tie this pin to V_{IN} if unused (for new chip).
FB	4	I	Feedback pin to set the output voltage with help of the feedback divider. See the Section 5.3 section for more information (for TPS763xx-Q1 Adjustable only).
GND	2	—	Ground
IN	1	I	Input supply pin. Use a capacitor with a value of 1µF or larger from this pin to ground. See the Section 5.3 and Section 7.1.3 section for more information.
NC	4	—	No connection (fixed-voltage option only).
OUT	5	O	Output of the regulator. Use a capacitor with a value of 2.2µF or larger from this pin to ground. See the Section 7.1.3 and section for more information.

(1) I = input, O = output

Note

The nominal output capacitance must be greater than 1µF. Throughout this document, the nominal derating on these capacitors is 50%. Make sure that the effective capacitance at the pin is greater than 1µF.

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5 Specifications

5.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
Voltage(2)	V_{IN} (for legacy chip)	–0.3	10	V
	V_{IN} (for new chip)	–0.3	18	
	V_{OUT}	–0.3	7	
	V_{FB} (for legacy chip)	–0.3	7	
	V_{FB} (for new chip)	–0.3	3	
	Voltage range at EN (for legacy chip)	–0.3	V_{IN} + 0.3	
	Voltage range at EN (for new chip)	–0.3	18	
Current	Maximum output current	Internally Limited		A
Temperature	Operating junction (T_J)	–40	150	°C
	Storage (T_{STG})	–65	150	

(1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the *Recommended Operating Conditions* but within the *Absolute Maximum Ratings*, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
(2) All voltages with respect to GND.

5.2 ESD Ratings

			VALUE (new chip)	VALUE (legacy chip)	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins(1)	±2000	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22C101, all pins(2)	±1000	±500	

(1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{IN}	Input voltage	2.7		10	V
EN	Enable voltage (for new chip)	0		10	
V_{OUT}	Output voltage (for new chip)	1.2		6.5	
I_{OUT}	Output current	0		150	mA
C_{OUT}	Output capacitance (for legacy chip)	4.7			µF
	Output capacitance (for new chip)	1	2.2	220	
C_{OUT} ESR	Output capacitor ESR (for legacy chip)	0.3		10	Ω
	Output capacitor ESR (for new chip)	0		1	
C_{IN}	Input capacitance (for new chip)		0.47		µF
T_J	Junction temperature (for legacy chip)	–40		125	°C
	Junction temperature (for new chip)	–40		150	°C

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5.4 Thermal Information

THERMAL METRIC[1]		Legacy chip	New chip	UNIT
		SOT-23 (DBV)	SOT-23 (DBV)	
		5 PINS	5 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	205.2	178.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	11.93	77.9	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	34.8	47.2	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	12.2	15.9	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	33.9	46.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application note.

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5.5 Electrical Characteristics

Specified at T_J = –40°C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.7V (whichever is greater), I_{OUT} = 1mA, EN = V_{IN}, C_{IN} = 1.0µF, C_{OUT} = 4.7µF (unless otherwise noted). Typical values are at T_J= 25ºC.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OUT}	Output voltage	TPS76301-Q1 (for legacy chip)	3.25V > V_{IN} ≥ 2.7V, 2.5V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 75mA, T_J = 25°C	0.98 × V_{OUT}	V_{OUT}	1.02 × V_{OUT}	V
			3.25V > V_{IN} ≥ 2.7V, 2.5V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 75mA	0.97 × V_{OUT}	V_{OUT}	1.03 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 100mA, T_J = 25°C	0.98 × V_{OUT}		1.02 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 100mA	0.97 × V_{OUT}		1.03 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 150mA, T_J = 25°C	0.975 × V_{OUT}		1.025 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 150mA	0.9625 × V_{OUT}		1.0375 × V_{OUT}	
		TPS76301-Q1 (for new chip)	3.25V > V_{IN} ≥ 2.7V, 2.5V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 75mA, T_J = 25°C	0.985 × V_{OUT}	V_{OUT}	1.015 × V_{OUT}	
			3.25V > V_{IN} ≥ 2.7V, 2.5V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 75mA	0.98 × V_{OUT}	V_{OUT}	1.02 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 100mA, T_J = 25°C	0.985 × V_{OUT}		1.015 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 100mA	0.98 × V_{OUT}		1.02 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 150mA, T_J = 25°C	0.985 × V_{OUT}		1.015 × V_{OUT}	
			V_{IN} ≥ 3.25V, 5.0V ≥ V_{OUT} ≥ 1.5V, I_{OUT} = 1mA to 150mA	0.98 × V_{OUT}		1.02 × V_{OUT}	
		TPS76316-Q1 (for legacy chip)	V_{IN} = 2.7V, I_{OUT} = 1mA to 75mA, T_J = 25°C	1.568	1.6	1.632	
			V_{IN} = 2.7V, I_{OUT} = 1mA to 75mA	1.552	1.6	1.648	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 100mA, T_J = 25°C	1.568	1.6	1.632	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 100mA	1.552	1.6	1.648	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 150mA, T_J = 25°C	1.56	1.6	1.64	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 150mA	1.536	1.6	1.664	
		TPS76318-Q1 (for legacy chip)	V_{IN} = 2.7V, I_{OUT} = 1mA to 75mA, T_J = 25°C	1.764	1.8	1.836	
			V_{IN} = 2.7V, I_{OUT} = 1mA to 75mA	1.746	1.8	1.854	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 100mA, T_J = 25°C	1.764	1.8	1.836	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 100mA	1.746	1.8	1.854	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 150mA, T_J = 25°C	1.755	1.8	1.845	
			V_{IN} = 3.25V, I_{OUT} = 1mA to 150mA	1.733	1.8	1.867	

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Specified at T_J = –40°C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.7V (whichever is greater), I_{OUT} = 1mA, EN = V_{IN}, C_{IN} = 1.0µF, C_{OUT} = 4.7µF (unless otherwise noted). Typical values are at T_J= 25ºC.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OUT}	Output voltage	TPS76325-Q1 (for legacy chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	2.45	2.5	2.55	V
			I_{OUT} = 1mA to 100mA	2.425	2.5	2.575	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	2.438	2.5	2.562	
			I_{OUT} = 1mA to 150mA	2.407	2.5	2.593	
		TPS76330-Q1 (for legacy chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	2.94	3	3.06	
			I_{OUT} = 1mA to 100mA	2.91	3	3.09	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	2.925	3	3.075	
			I_{OUT} = 1mA to 150mA	2.888	3	3.112	
		TPS76333-Q1 (for legacy chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	3.234	3.3	3.366	
			I_{OUT} = 1mA to 100mA	3.201	3.3	3.399	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	3.218	3.3	3.382	
			I_{OUT} = 1mA to 150mA	3.177	3.3	3.423	
		TPS76333-Q1 (for new chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	3.234	3.3	3.366	
			I_{OUT} = 1mA to 100mA	3.20925	3.3	3.39075	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	3.234	3.3	3.366	
			I_{OUT} = 1mA to 150mA	3.20925	3.3	3.39075	
		TPS76350-Q1 (for legacy chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	4.875	5	5.125	
			I_{OUT} = 1mA to 100mA	4.825	5	5.175	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	4.85	5	5.15	
			I_{OUT} = 1mA to 150mA	4.8	5	5.2	
		TPS76350-Q1 (for new chip)	I_{OUT} = 1mA to 100mA, T_J = 25°C	4.9	5	5.1	
			I_{OUT} = 1mA to 100mA	4.8625	5	5.1375	
			I_{OUT} = 1mA to 150mA, T_J = 25°C	4.9	5	5.1	
			I_{OUT} = 1mA to 150mA	4.8625	5	5.1375	
I_Q	Quiescent current (GND current)	For legacy chip	I_{OUT} = 1mA to 150mA, T_J = 25°C		85	100	µA
			I_{OUT} = 1mA to 150mA			140	
		For new chip	I_{OUT} = 0mA		65	125	
			I_{OUT} = 1mA to 150mA, T_J = 25°C		765	890	
			I_{OUT} = 1mA to 150mA			1120	
$\Delta V_{OUT(\Delta VOUT)}$	Output voltage line regulation ($\Delta V_{OUT}/V_{OUT}$)	For legacy chip	$V_{OUT(NOM)}$ +1.0V ≤ V_{IN} ≤ 10V, V_{IN} ≥ 3.5V, T_J = 25°C		0.04	0.07	%/V
			$V_{OUT(NOM)}$ +1.0V ≤ V_{IN} ≤ 10V, V_{IN} ≥ 3.5V			0.1	
		For new chip	$V_{OUT(NOM)}$ +1.0V ≤ V_{IN} ≤ 10V, V_{IN} ≥ 3.5V, T_J = 25°C			0.01	
			$V_{OUT(NOM)}$ +1.0V ≤ V_{IN} ≤ 10V, V_{IN} ≥ 3.5V			0.01	
V_n	Output noise voltage	For legacy chip	BW = 300Hz to 50kHz, V_{OUT} = 3.3V , C_{OUT} = 10µF		140		μV_{RMS}
		For new chip	BW = 300Hz to 50kHz, V_{OUT} = 3.3V , C_{OUT} = 4.7µF		165		
I_{CL}	Output current limit	For legacy chip	V_{OUT} = 0V	0.5	0.8	1.5	A
		For new chip			0.8	1.05	

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Specified at T_J = –40°C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.7V (whichever is greater), I_{OUT} = 1mA, EN = V_{IN}, C_{IN} = 1.0µF, C_{OUT} = 4.7µF (unless otherwise noted). Typical values are at T_J= 25ºC.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{STANDBY}$	Standby current	For legacy chip	EN < 0.5V, T_J = 25°C		0.5	1	µA
			EN < 0.5V			2	
		For new chip	EN < 0.15V, T_J = 25°C		1.25		
			EN < 0.15V		1.12	3.75	
EN	High level enable input voltage	For legacy chip			1.4	2	V
	Low level enable input voltage			0.5	1.2		
	High level enable input voltage	For new chip			0.85	1.6	
	Low level enable input voltage			0.15	0.72		
PSRR	Power-supply ripple rejection	For legacy chip	C_{OUT} = 10µF, f = 1kHz, T_J = 25 °C		60		dB
		For new chip	C_{OUT} = 4.7µF, f = 1kHz, T_J = 25 °C		58		
I_{EN}	Input current (EN)	For legacy chip	EN = 0V		–0.01	–0.5	µA
			EN = V_{IN}		–0.01	–0.5	
		For new chip	EN = 0V		–0.35	–0.7	
			EN = V_{IN}		0.008	0.8	
V_{DO}	Dropout voltage	TPS76325-Q1 (for legacy chip)	I_{OUT} = 0mA, T_J = 25°C		0.2		mV
			I_{OUT} = 1mA, T_J = 25°C		3		
			I_{OUT} = 50mA, T_J = 25°C		120	150	
			I_{OUT} = 50mA			200	
			I_{OUT} = 75mA, T_J = 25°C		180	225	
			I_{OUT} = 75mA			300	
			I_{OUT} = 100mA, T_J = 25°C		240	300	
			I_{OUT} = 100mA			400	
			I_{OUT} = 150mA, T_J = 25°C		360	450	
			I_{OUT} = 150mA			600	
		TPS76333-Q1 (for legacy chip)	I_{OUT} = 0mA, T_J = 25°C		0.2		
			I_{OUT} = 1mA, T_J = 25°C		3		
			I_{OUT} = 50mA, T_J = 25°C		100	125	
			I_{OUT} = 50mA			166	
			I_{OUT} = 75mA, T_J = 25°C		150	188	
			I_{OUT} = 75mA			250	
			I_{OUT} = 100mA, T_J = 25°C		200	250	
			I_{OUT} = 100mA			333	
			I_{OUT} = 150mA, T_J = 25°C		300	375	
			I_{OUT} = 150mA			500	

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Specified at T_J = –40°C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.7V (whichever is greater), I_{OUT} = 1mA, EN = V_{IN}, C_{IN} = 1.0µF, C_{OUT} = 4.7µF (unless otherwise noted). Typical values are at T_J= 25ºC.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{DO}	Dropout voltage	TPS76350-Q1 (for legacy chip)	I_{OUT} = 0mA, T_J = 25℃		0.2		mV
			I_{OUT} = 1mA, T_J = 25℃		2		
			I_{OUT} = 50mA, T_J = 25℃		60	75	
			I_{OUT} = 50mA			100	
			I_{OUT} = 75mA, T_J = 25℃		90	113	
			I_{OUT} = 75mA			150	
			I_{OUT} = 100mA, T_J = 25℃		120	150	
			I_{OUT} = 100mA			200	
			I_{OUT} = 150mA, T_J = 25℃		180	225	
			I_{OUT} = 150mA			300	
		TPS763xx-Q1 (for new chip)	I_{OUT} = 0mA, T_J = 25℃		1		
			I_{OUT} = 1mA, T_J = 25℃		10.5		
			I_{OUT} = 50mA, T_J = 25℃		125	150	
			I_{OUT} = 50mA			205	
			I_{OUT} = 75mA, T_J = 25℃		135	155	
			I_{OUT} = 75mA			210	
			I_{OUT} = 100mA, T_J = 25℃		145	165	
			I_{OUT} = 100mA			225	
			I_{OUT} = 150mA, T_J = 25℃		175	195	
			I_{OUT} = 150mA			260	

5.6 Typical Characteristics

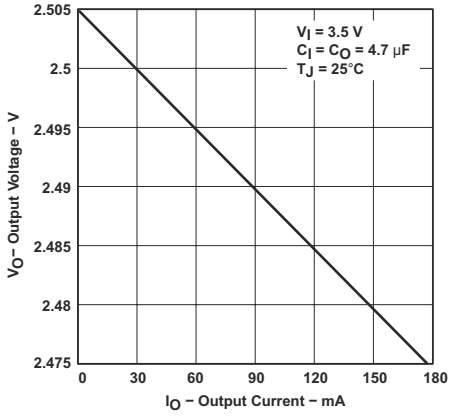


Figure 5-1. TPS76325-Q1 Output Voltage vs Output Current (Legacy Chip)

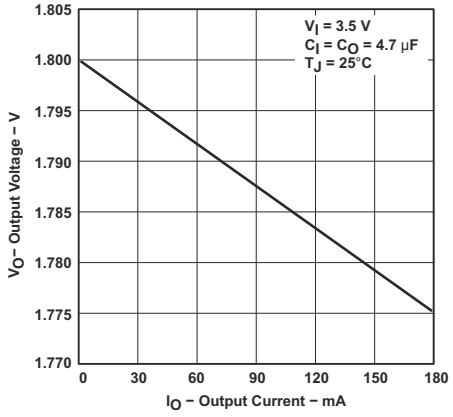


Figure 5-2. TPS76318-Q1 Output Voltage vs Output Current (Legacy Chip)

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5.6 Typical Characteristics (continued)

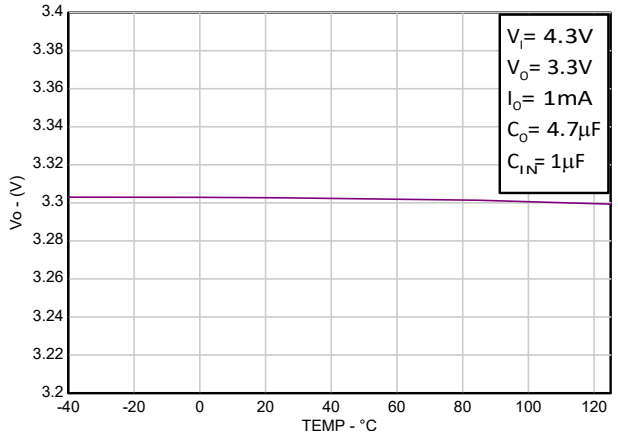


Figure 5-3. TPS76333-Q1 Output Voltage vs Output Current (new chip)

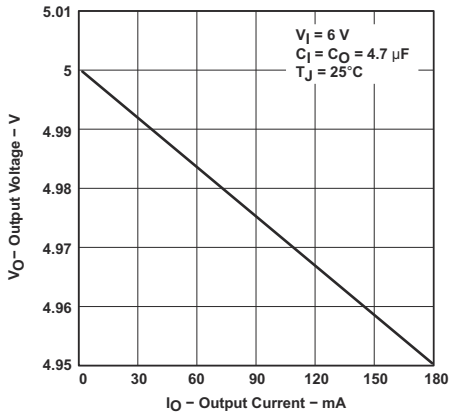


Figure 5-4. TPS76350-Q1 Output Voltage vs Output Current (Legacy Chip)

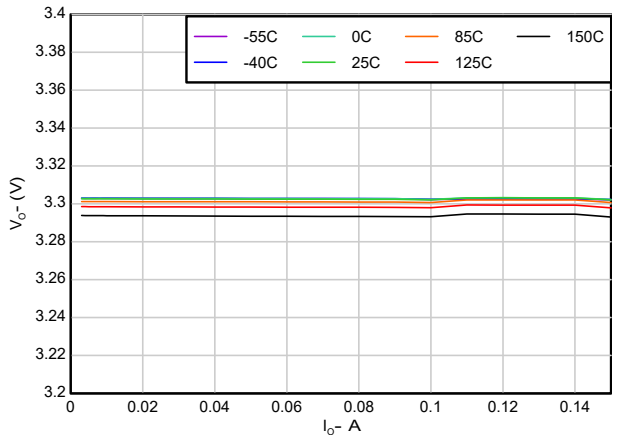


Figure 5-5. TPS76333-Q1 Output Voltage vs Output Current (new chip)

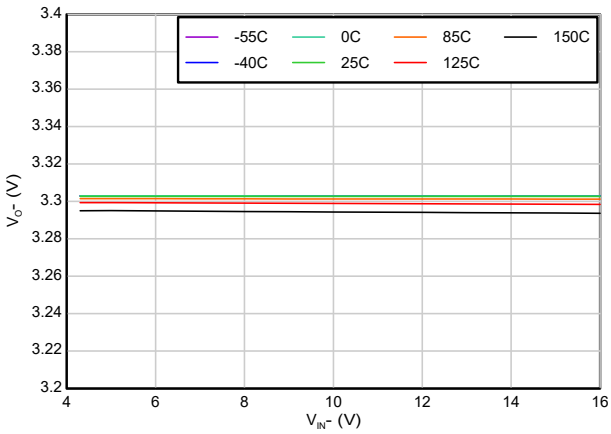


Figure 5-6. TPS76333-Q1 Output Voltage vs Input Voltage (new chip)

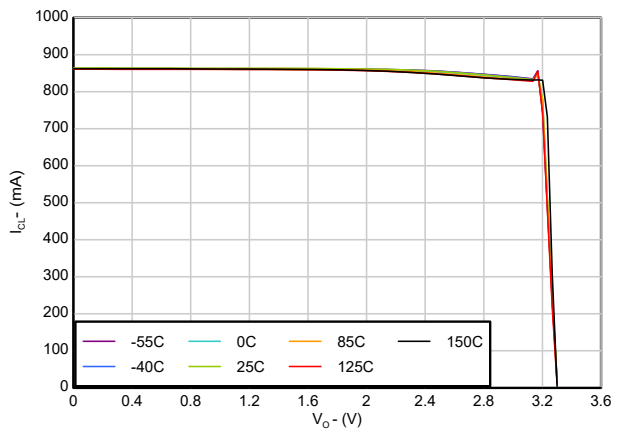


Figure 5-7. TPS76333-Q1 Short-circuit current vs Output voltage (new chip)

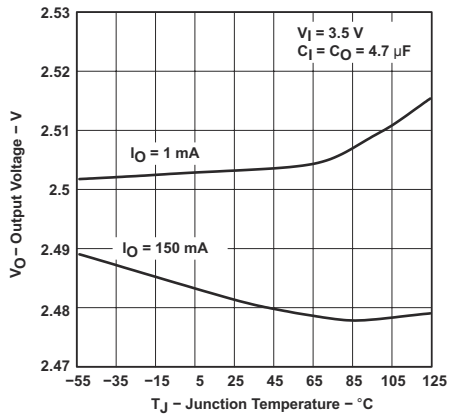


Figure 5-8. TPS76325-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

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5.6 Typical Characteristics (continued)

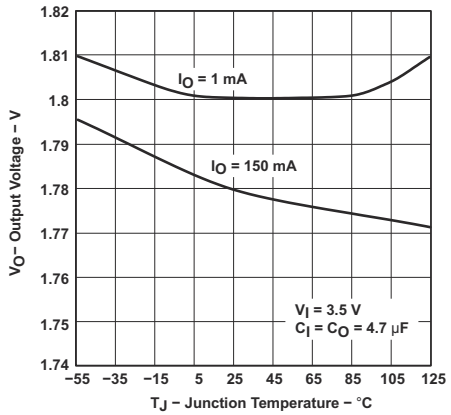


Figure 5-9. TPS76318-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

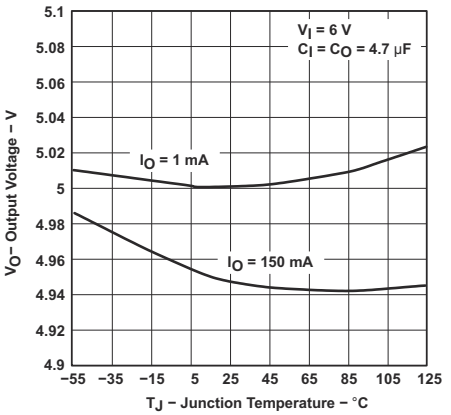


Figure 5-10. TPS76350-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

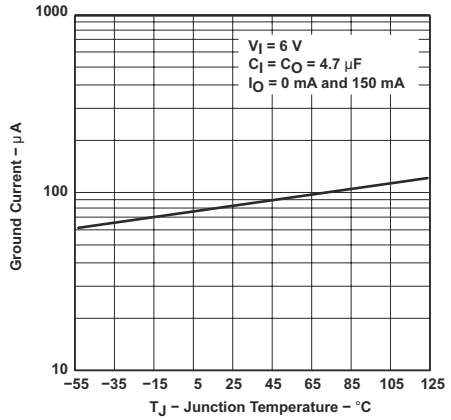


Figure 5-11. TPS76350-Q1 Ground Current vs Free-Air Temperature (Legacy Chip)

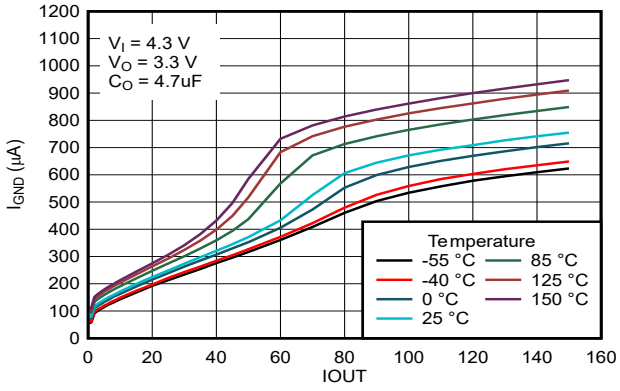


Figure 5-12. Ground Pin Current vs Load Current (new chip)

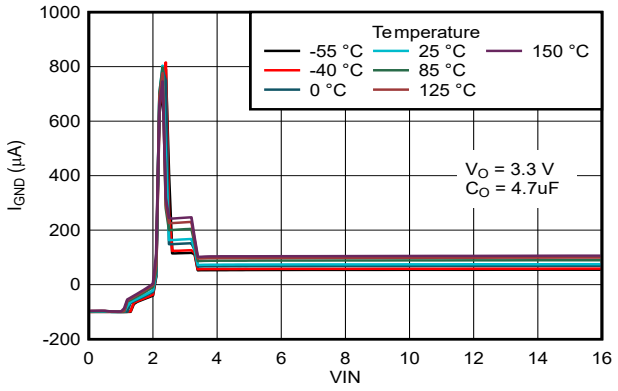


Figure 5-13. Input Current vs Input Voltage (new chip)

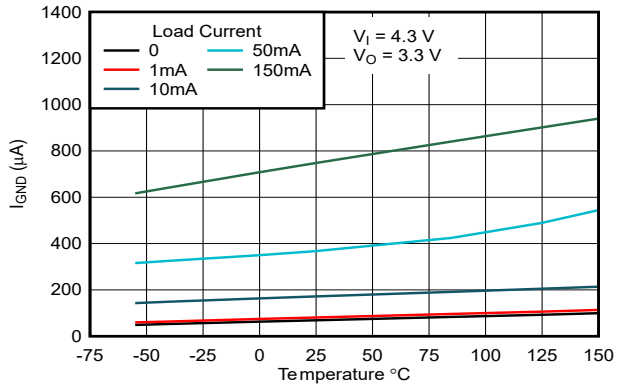


Figure 5-14. Ground-Pin Current vs Temperature (new chip)

5.6 Typical Characteristics (continued)

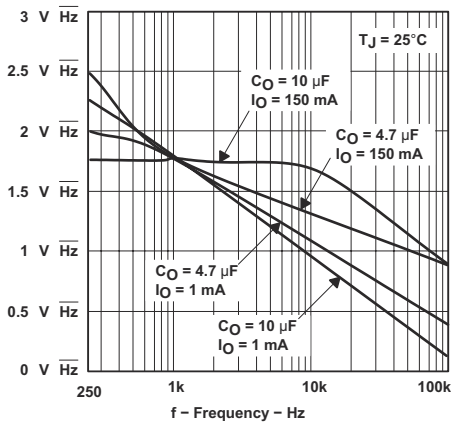


Figure 5-15. Output Noise vs Frequency (Legacy Chip)

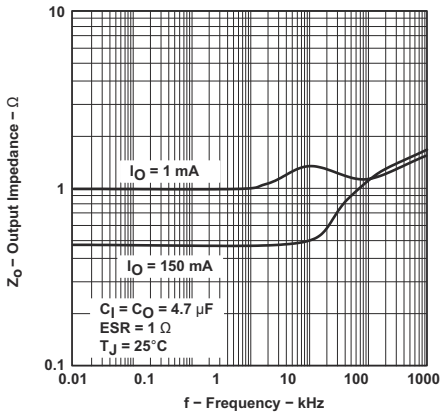


Figure 5-16. Output Impedance vs Frequency (Legacy Chip)

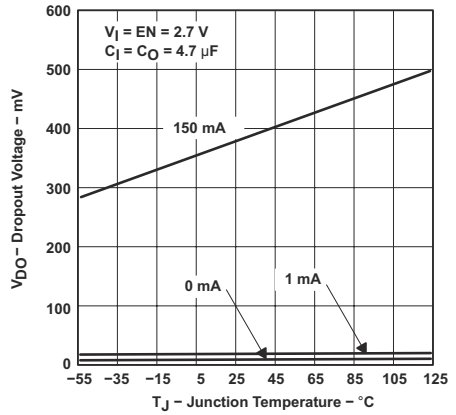


Figure 5-17. TPS76325-Q1 Dropout Voltage vs Free-Air Temperature (Legacy Chip)

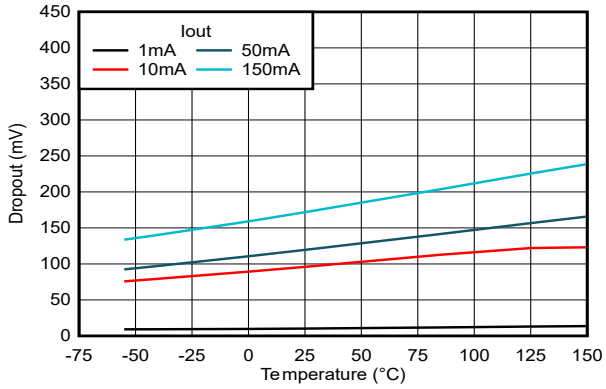


Figure 5-18. TPS76333-Q1 Dropout Voltage vs Temperature (new chip)

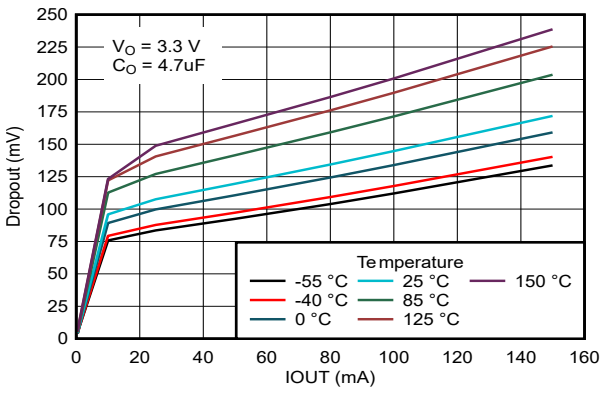


Figure 5-19. TPS76333-Q1 Dropout Voltage vs Load Current (new chip)

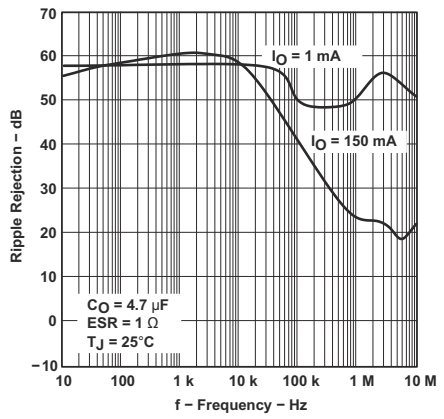


Figure 5-20. TPS76325-Q1 Ripple Rejection vs Frequency (Legacy Chip)

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5.6 Typical Characteristics (continued)

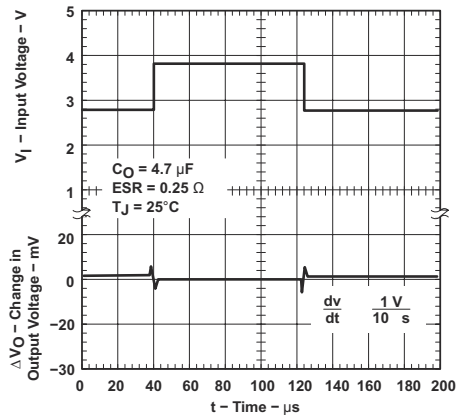


Figure 5-21. TPS76318-Q1 Line Transient Response (Legacy Chip)

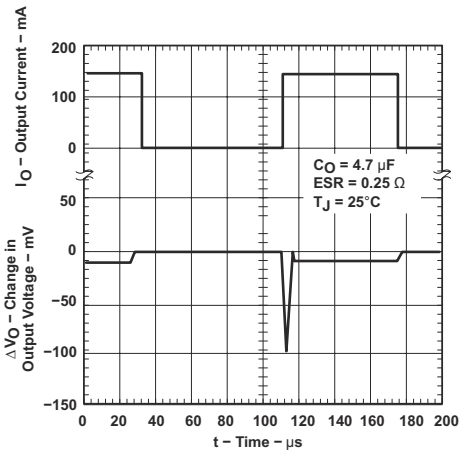


Figure 5-22. TPS76318-Q1 Load Transient Response (Legacy Chip)

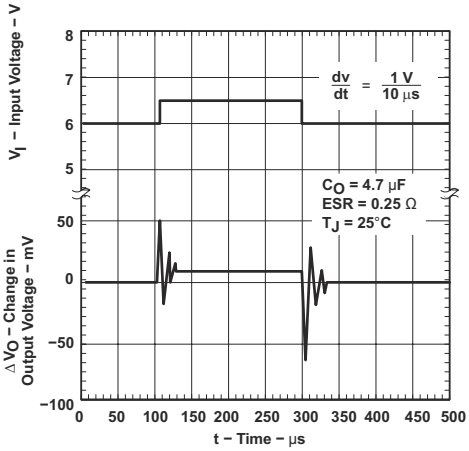


Figure 5-23. TPS76350-Q1 Line Transient Response (Legacy Chip)

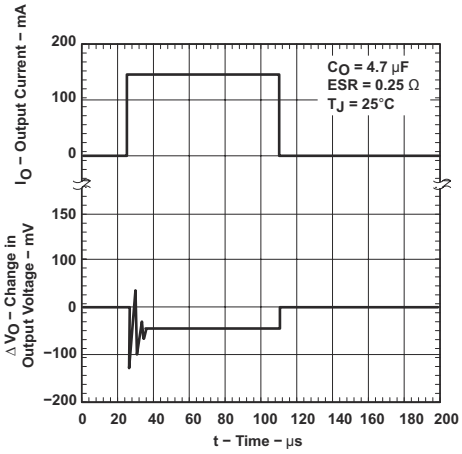
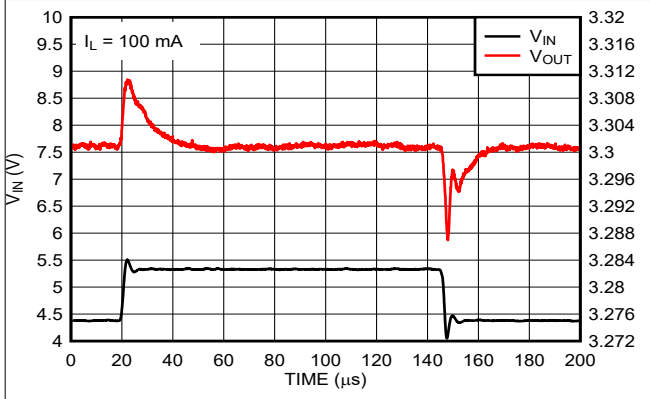
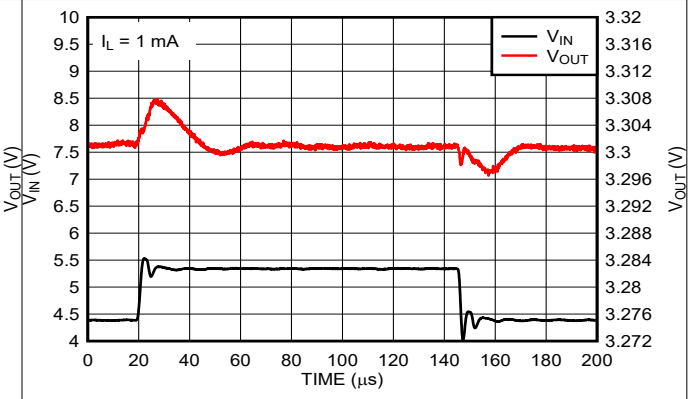


Figure 5-24. TPS76350-Q1 Load Transient Response (Legacy Chip)



V_{OUT} = 3.3V, I_L = 100mA

Figure 5-25. Line Transient Response (New Chip)

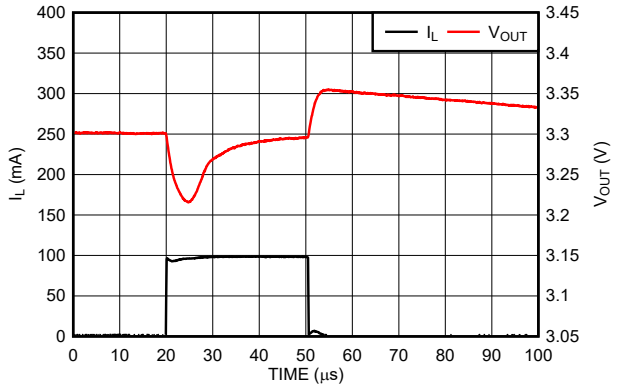


V_{OUT} = 3.3V, I_L = 1mA

Figure 5-26. Line Transient Response (New Chip)

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5.6 Typical Characteristics (continued)



V_{OUT} = 3.3V, C_{OUT} = 2.2µF

Figure 5-27. Load Transient Response (New Chip)

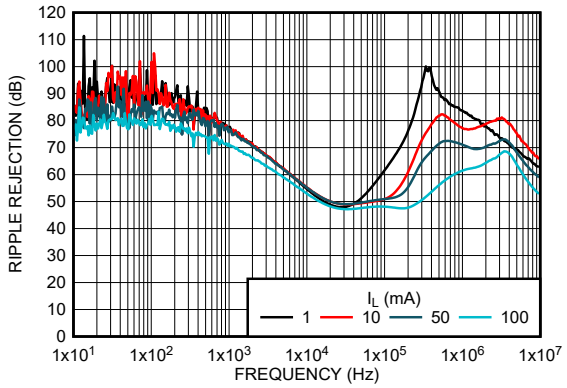


Figure 5-28. Ripple Rejection versus Load Current (I_L) and Frequency (New Chip)

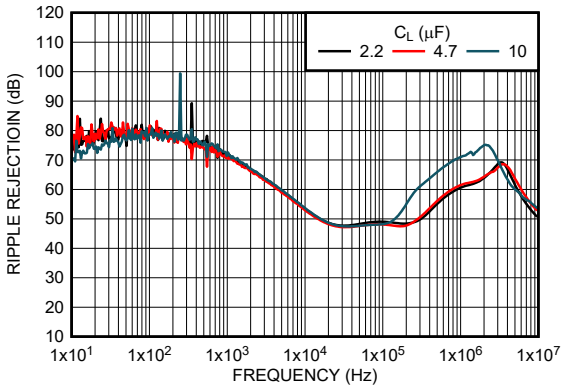


Figure 5-29. Ripple Rejection versus Output Capacitor (C_L) and Frequency (New Chip)

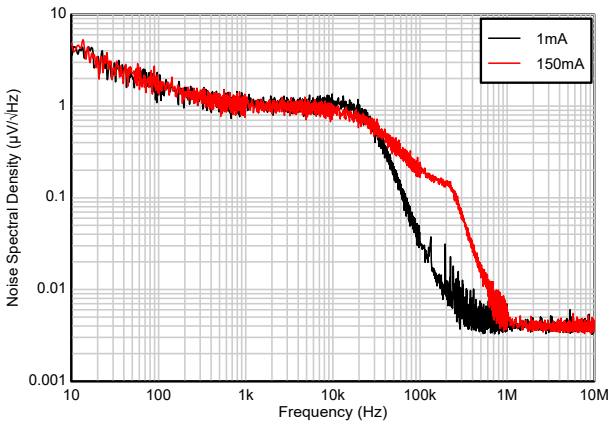
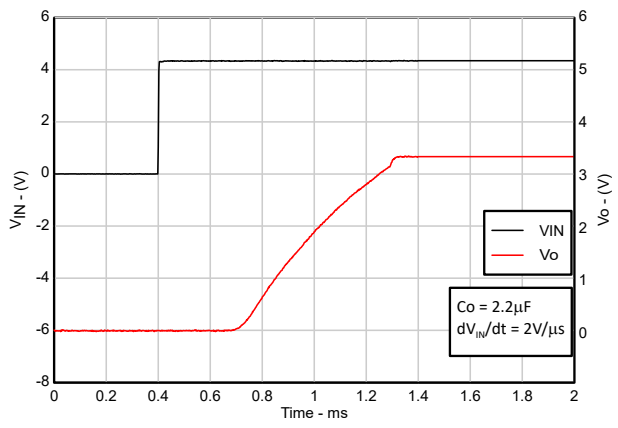
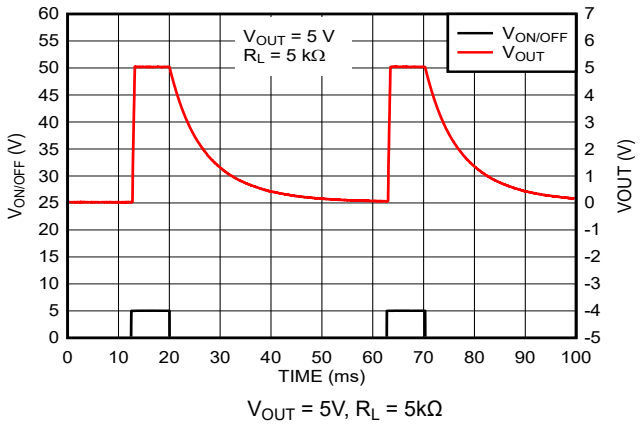


Figure 5-30. Output Noise Density versus Load Current (I_L) Frequency (New Chip)



V_{OUT} = 3.3V, R_L = 3.3kΩ

Figure 5-31. Turn-on Waveform (New Chip)

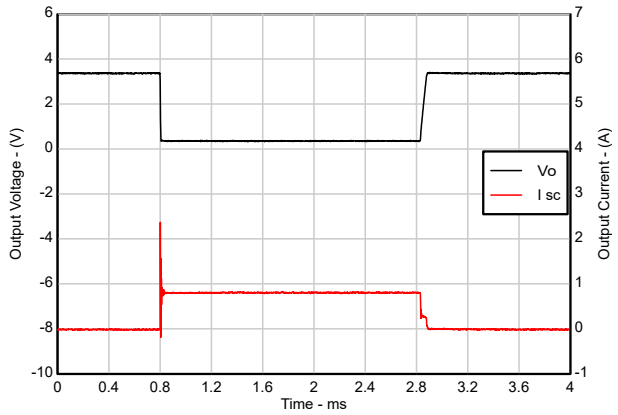


V_{OUT} = 5V, R_L = 5kΩ

Figure 5-32. Turn-off Waveform (New Chip)

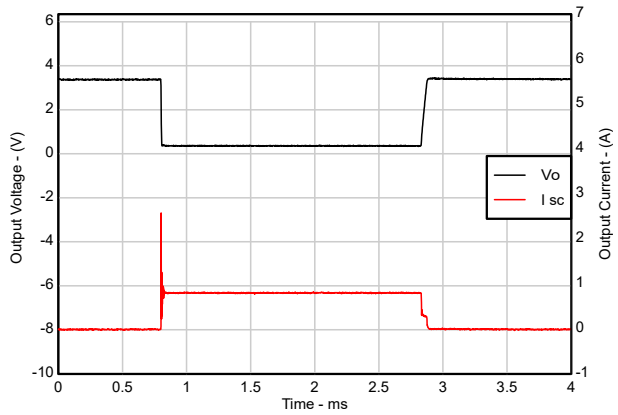
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5.6 Typical Characteristics (continued)



V_{IN} = 4.3V

Figure 5-33. Short Circuit Current versus Time (New Chip)



V_{IN} = 10.0V

Figure 5-34. Short Circuit Current versus Time (New Chip)

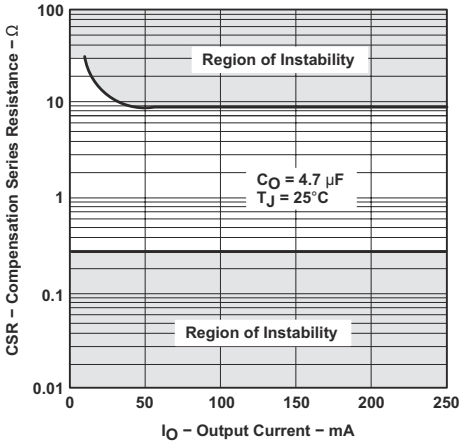


Figure 5-35. Typical Regions of Stability Compensation Series Resistance (CSR) vs Output Current (Legacy Chip)

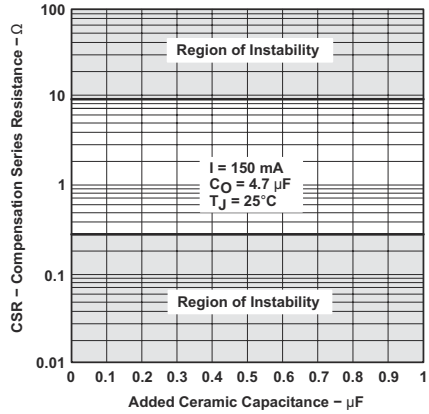


Figure 5-36. Typical Regions of Stability Compensation Series Resistance (CSR) vs Added Ceramic Capacitance (Legacy Chip)

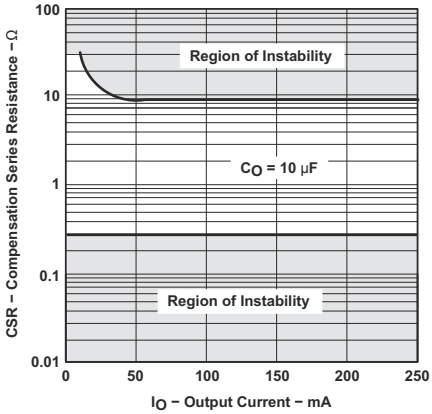


Figure 5-37. Typical Regions of Stability Compensation Series Resistance (CSR) vs Output Current (Legacy Chip)

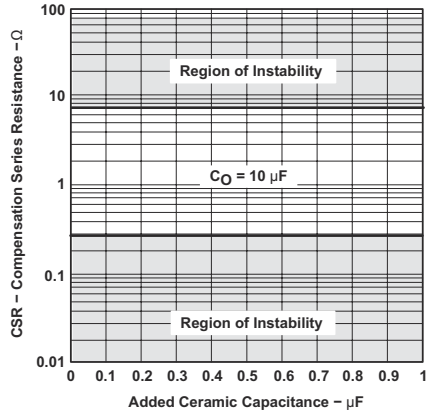


Figure 5-38. Typical Regions of Stability Compensation Series Resistance (CSR) vs Added Ceramic (Legacy Chip)

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6 Detailed Description

6.1 Overview

The TPS763xx-Q1 family of low-dropout (LDO) linear voltage regulators supports wide input voltage range of 2.7V to 10V and up to 150mA of load current. The output range is from 1.6V to 5.0V for the fixed version, and from 1.6V to 6.5V for the adjustable version.

The TPS763xx-Q1 has a ±1.5% output accuracy that is required for powering digital loads with tight supply requirements. The TPS763xx-Q1 (new chip) has an internal soft start mechanism that provides a uniform start-up with controlled inrush current. This LDO also has over-current and thermal protection during a load-short or fault condition on the output for better reliability.

6.2 Functional Block Diagram

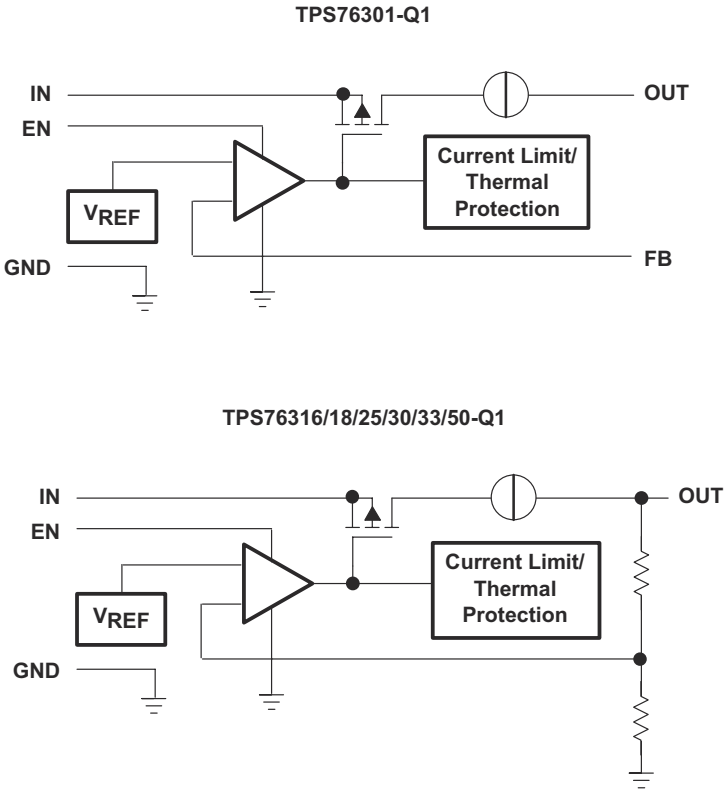


Figure 6-1. Functional Block Diagram (for Legacy Chip)

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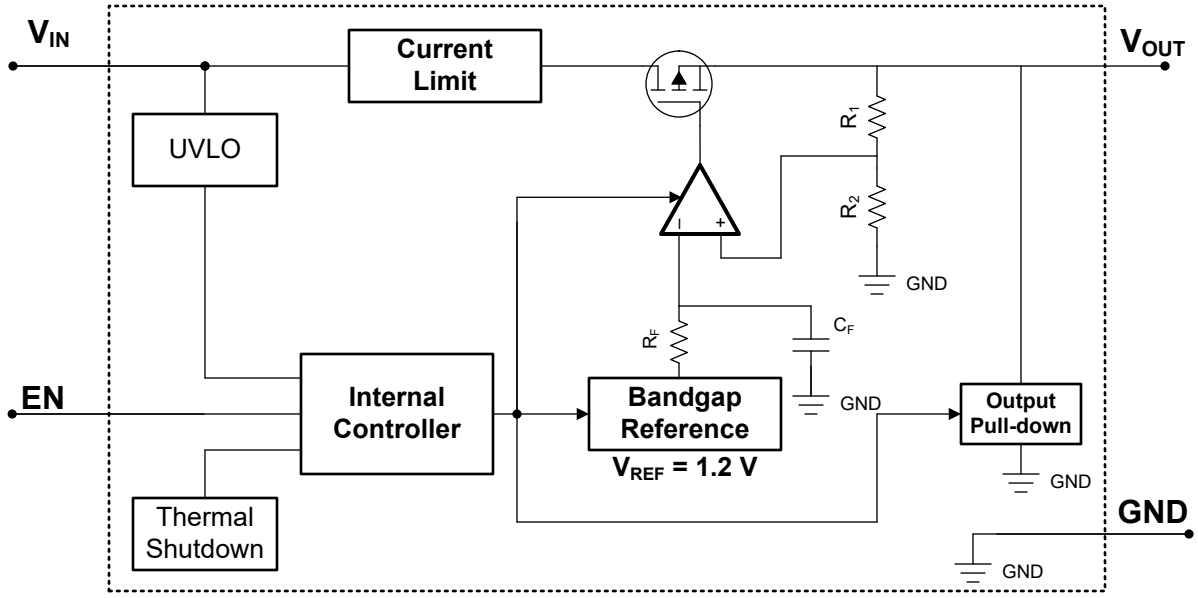


Figure 6-2. Functional Block Diagram (Fixed, New Chip)

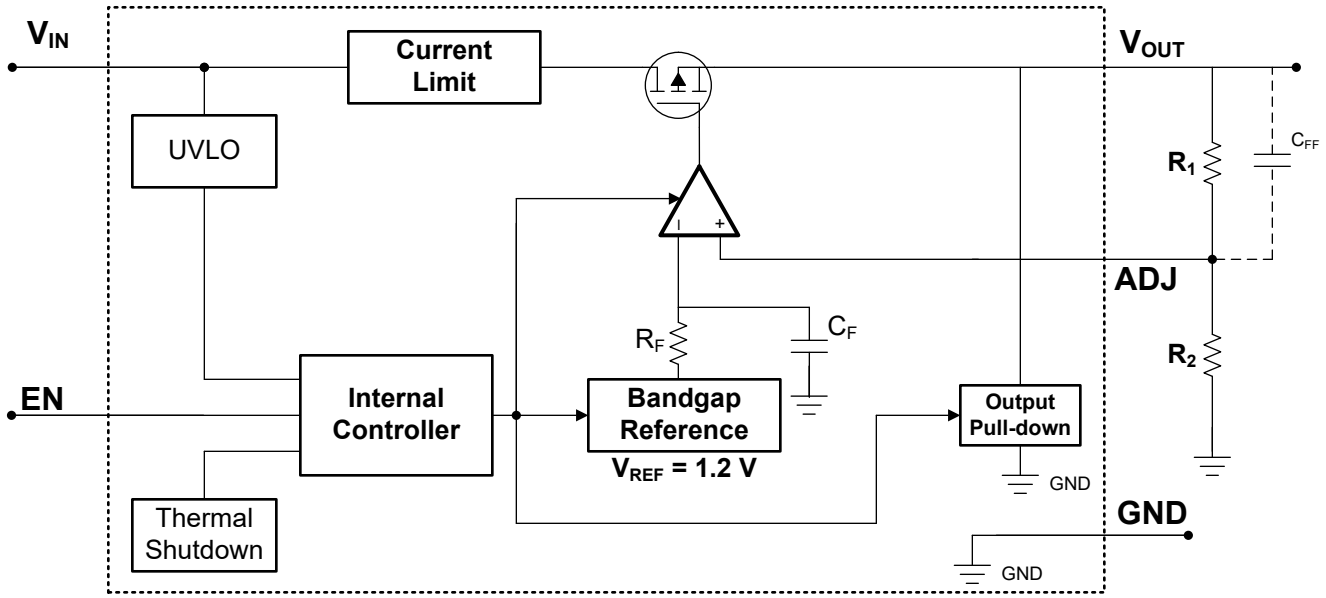


Figure 6-3. Functional Block Diagram (Adjustable, New Chip)

6.3 Feature Description

6.3.1 Output Enable

The EN pin for the device is an active-high pin. The output voltage is enabled when the voltage of the EN pin is greater than the high-level input voltage of the EN pin and disabled with the EN pin voltage is less than the low-level input voltage of the EN pin. If independent control of the output voltage is not needed, connect the EN pin to the input of the device.

The device has an internal pulldown circuit that activates when the device is disabled by pulling the EN pin voltage lower than the low-level input voltage of the EN pin to actively discharge the output voltage.

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6.3.2 Dropout Voltage

Dropout voltage (V_{DO}) is defined as the input voltage minus the output voltage ($V_{IN} – V_{OUT}$) at the rated output current (I_{RATED}), where the pass transistor is fully on. I_{RATED} is the maximum I_{OUT} listed in the Recommended Operating Conditions table. The pass transistor is in the ohmic or triode region of operation, and acts as a switch. The dropout voltage indirectly specifies a minimum input voltage greater than the nominal programmed output voltage at which the output voltage is expected to stay in regulation. If the input voltage falls to less than the nominal output regulation, then the output voltage falls as well.

For a CMOS regulator, the dropout voltage is determined by the drain-source on-state resistance ($R_{DS(ON)}$) of the pass transistor. Therefore, if the linear regulator operates at less than the rated current, the dropout voltage for that current scales accordingly. The following equation calculates the $R_{DS(ON)}$ of the device.

$$R_{DS(ON)} = \frac{V_{DO}}{I_{RATED}} \quad (1)$$

6.3.3 Current Limit

For the legacy chip, the internal current limit circuit protects the LDO against high-load current faults or shorting events. The LDO is not designed to operate in a steady-state current limit. During a current-limit event, the LDO sources constant current. Therefore, the output voltage falls when load impedance decreases. If a current limit occurs and the resulting output voltage is low, excessive power is potentially dissipated across the LDO, resulting in a thermal shutdown of the output. A foldback feature limits the short-circuit current to protect the regulator from damage under all load conditions. If V_{OUT} is forced below 0V before EN goes high and the load current required exceeds the foldback current limit, the device potentially does not start up correctly.

For the new chip, the device has an internal current limit circuit that protects the regulator during transient high-load current faults or shorting events. The current limit is a brick-wall scheme. In a high-load current fault, the brick-wall scheme limits the output current to the current limit (I_{CL}). I_{CL} is listed in the Electrical Characteristics table.

The output voltage is not regulated when the device is in current limit. When a current limit event occurs, the device begins to heat up because of the increase in power dissipation. When the device is in brick-wall current limit, the pass transistor dissipates power [($V_{IN} – V_{OUT}$) × I_{CL}]. If thermal shutdown is triggered, the device turns off. After the device cools down, the internal thermal shutdown circuit turns the device back on. If the output current fault condition continues, the device cycles between current limit and thermal shutdown. For more information on current limits, see the *Know Your Limits* application note.

Figure 6-4 shows a diagram of the current limit.

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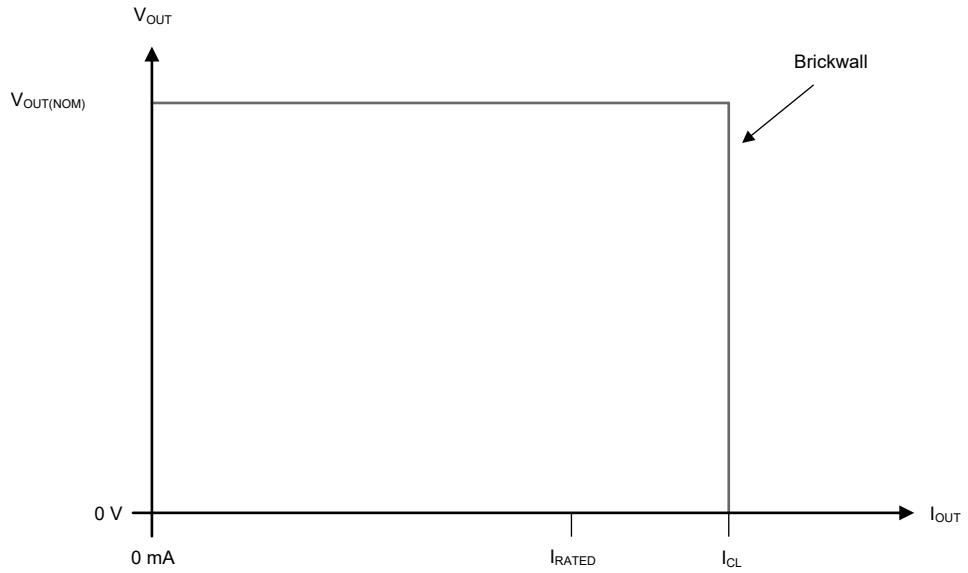


Figure 6-4. Current Limit

6.3.4 Output Pulldown

The new chip has an output pulldown circuit. The output pulldown activates in the following conditions:

- When the device is disabled ($V_{EN} < V_{EN(LOW)}$)
- If 1.0V < V_{IN} < 2.7V (for new chip)

Do not rely on the output pulldown circuit for discharging a large amount of output capacitance after the input supply has collapsed because reverse current can flow from the output to the input. This reverse current flow can cause damage to the device. See the *Section 7.1.4* section for more details.

6.3.5 Thermal Shutdown

The device contains a thermal shutdown protection circuit to disable the device when the junction temperature (T_J) of the pass transistor rises to $T_{SD(shutdown)}$ (typical). Thermal shutdown hysteresis makes sure the device resets (turns on) when the temperature falls to $T_{SD(reset)}$ (typical).

The thermal time-constant of the semiconductor die is fairly short, thus the device can cycle on and off when thermal shutdown is reached until power dissipation is reduced. Power dissipation during start up can be high from large V_{IN} – V_{OUT} voltage drops across the device or from high inrush currents charging large output capacitors. Under some conditions, the thermal shutdown protection disables the device before start up completes.

For reliable operation, limit the junction temperature to the maximum listed in the Recommended Operating Conditions table. Operation above this maximum temperature causes the device to exceed operational specifications. Although the internal protection circuitry of the device is designed to protect against thermal overall conditions, this circuitry is not intended to replace proper heat sinking. Continuously running the device into thermal shutdown or above the maximum recommended junction temperature reduces long-term reliability.

6.4 Device Functional Modes

6.4.1 Device Functional Mode Comparison

Table 6-1 shows the conditions that lead to the different modes of operation. See the *Section 5.5* table for parameter values.

Table 6-1. Device Functional Mode Comparison

OPERATING MODE	PARAMETER			
	V_{IN}	$V_{ON/\overline{OFF}}$	I_{OUT}	T_J
Normal operation	$V_{IN} > V_{OUT(nom)} + V_{DO}$ and $V_{IN} > V_{IN(min)}$	$V_{EN} > V_{EN(HI)}$	$I_{OUT} < I_{OUT(max)}$	$T_J < T_{SD(shutdown)}$
Dropout operation	$V_{IN(min)} < V_{IN} < V_{OUT(nom)} + V_{DO}$	$V_{EN} > V_{EN(HI)}$	$I_{OUT} < I_{OUT(max)}$	$T_J < T_{SD(shutdown)}$
Disabled (any true condition disables the device)	$V_{IN} < 2.7V$	$V_{EN} < V_{EN(LOW)}$	Not applicable	$T_J > T_{SD(shutdown)}$

6.4.2 Normal Operation

The device regulates to the nominal output voltage when the following conditions are met:

- The input voltage is greater than the nominal output voltage plus the dropout voltage ($V_{OUT(nom)} + V_{DO}$)
- The output current is less than the current limit ($I_{OUT} < I_{CL}$)
- The device junction temperature is less than the thermal shutdown temperature ($T_J < T_{SD}$)
- The EN voltage has previously exceeded the EN rising threshold voltage and has not yet decreased to less than the enable falling threshold

6.4.3 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, but all other conditions are met for normal operation, the device operates in dropout mode. In this mode, the output voltage tracks the input voltage. During this mode, the transient performance of the device becomes significantly degraded because the pass transistor is in the ohmic or triode region, and acts as a switch. Line or load transients in dropout can result in large output voltage deviations.

When the device is in a steady dropout state (defined as when the device is in dropout, $V_{IN} < V_{OUT(NOM)} + V_{DO}$, directly after being in a normal regulation state, but *not* during start up), the pass transistor is driven into the ohmic or triode region. When the input voltage returns to a value greater than or equal to the nominal output voltage plus the dropout voltage ($V_{OUT(NOM)} + V_{DO}$), the output voltage can overshoot for a short period of time while the device pulls the pass transistor back into the linear region.

6.4.4 Disabled

The output of the device can be shutdown by forcing the voltage of the EN pin to less than the maximum EN pin low-level input voltage (see the Electrical Characteristics table). When disabled, the pass transistor is turned off, internal circuits are shutdown, and the output voltage is actively discharged to ground by an internal discharge circuit from the output to ground.

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7 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Application Information

7.1.1 Adjustable Device Feedback Resistors

The adjustable-version device requires external feedback divider resistors to set the output voltage. V_{OUT} is set using the feedback divider resistors, R_1 and R_2, according to the following equation:

$$V_{OUT} = V_{FB} \times \left[\frac{1 + R_1}{R_2}\right] \quad (2)$$

7.1.2 Recommended Capacitor Types

7.1.2.1 Recommended Capacitors (Legacy Chip)

The TPS763xx-Q1 is designed to work with an output capacitor connected between OUT and GND to stabilize the internal loop control. The minimum recommended capacitance value is 4.7µF and the ESR (equivalent series resistance) must be between 0.3Ω and 10Ω. Capacitor values of 4.7µF or larger are acceptable, provided the ESR is less than 10Ω. Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all good choice for the stable operation, provided these capacitors meet the requirements described above.

One disadvantage of ceramic capacitors is that the capacitance varies with temperature. Most large-value ceramic capacitors (≥ 2.2µF) are manufactured with the Z5U or Y5V temperature characteristic. Thus, resulting in the capacitance dropping by more than 50% as the temperature goes from 25°C to 85°C.

This capacitance drop potentially cause problems if a 4.7µF capacitor is used on the output because that capacitor goes down to approximately 4.7µF at high ambient temperatures. Such low capacitance causes the TPS763xx-Q1 to oscillate. If Z5U or Y5V capacitors are used on the output, adhere to a minimum capacitance value of 4.7µF.

A better choice for temperature coefficient in ceramic capacitors is X7R, which holds the capacitance within ±15%. Unfortunately, the larger values of capacitance are not offered by all manufacturers in the X7R dielectric.

Tantalum capacitors are less desirable than ceramics for use as output capacitors. These components are more expensive when comparing equivalent capacitance and voltage ratings in the 1µF to 4.7µF range.

Another important consideration is that tantalum capacitors have higher ESR values than equivalent size ceramics. Which means that although a tantalum capacitor potentially has an ESR value within the stable range, the capacitor is larger in capacitance. Thus, the tantalum capacitor is bigger and more costly than a ceramic capacitor with the same ESR value.

The ESR of a typical tantalum increases by approximately 2:1 as the temperature goes from 25°C down to –40°C, so some allow for guard band.

7.1.2.2 Recommended Capacitors (New Chip)

The new chip is designed to be stable using low equivalent series resistance (ESR) ceramic capacitors at the input and output. Multilayer ceramic capacitors have become the industry standard for these types of applications and are recommended, but use good judgment. Ceramic capacitors that employ X7R-, X5R-, and C0G-rated dielectric materials provide relatively good capacitive stability across temperature. However, using Y5V-rated capacitors is discouraged because of large variations in capacitance.

Maximum supported ESR range across complete temperature (−40°C to +150°C) and load current range (0mA-150mA) is less than 1Ω. If in an existing implementation where different type of capacitors with higher

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ESR are used, use a low-ESR, 100nF MLCC capacitor. Place this capacitor as close as possible to the device output pin (V_{OUT}).

Regardless of the ceramic capacitor type selected, the effective capacitance varies with operating voltage and temperature. Generally, expect the effective capacitance to decrease by as much as 50%. The input and output capacitors listed in the Recommended Operating Conditions table account for an effective capacitance of approximately 50% of the nominal value.

7.1.3 Input and Output Capacitor Requirements

7.1.3.1 Input Capacitor Requirements

For legacy chip, although not required, a 0.047µF or larger ceramic bypass input capacitor, connected between IN and GND and located close to the TPS763xx-Q1, is recommended to improve transient response and noise rejection. A higher-value electrolytic input capacitor can be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

Note

Tantalum capacitors can suffer catastrophic failure because of surge current when connected to a low-impedance source of power (such as a battery or very large capacitor). If a tantalum capacitor is used at the input, contact the manufacturer to make sure the capacitor has a surge current rating sufficient for the application.

There are no requirements for ESR on the input capacitor, but consider tolerance and temperature coefficient when selecting the capacitor to make sure the capacitance is ≥ 1µF over the entire operating temperature range.

For new chip, although an input capacitor is not required for stability, good analog design practice is to connect a capacitor from IN to GND. This capacitor counteracts reactive input sources and improves transient response, input ripple, and PSRR. Use an input capacitor if the source impedance is more than 0.5Ω. A higher value capacitor can be necessary if large, fast rise-time load or line transients are anticipated or if the device is located several inches from the input power source.

7.1.3.2 Output Capacitor Requirements

Like all low dropout regulators, the TPS763xx-Q1 requires an output capacitor connected between OUT and GND to stabilize the internal loop control. The minimum recommended capacitance value is 4.7µF and the ESR (equivalent series resistance) must be between 0.3Ω and 10Ω. Capacitor values of 4.7µF or larger are acceptable, provided the ESR is less than 10Ω. Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all excellent choices, provided the requirements described above are met. Most of the commercially available 4.7µF surface-mount solid tantalum capacitors, including devices from Sprague, Kemet, and Nichico, meet the ESR requirements stated above.

Table 7-1. Capacitor Selection

PART NO.	MANUFACTURER	VALUE	MAX ESR	SIZE (H × L × W)
T494B475K016AS	KEMET	4.7µF	1.5Ω	1.9 × 3.5 × 2.8
195D106x0016x2T	SPRAGUE	10µF	1.5Ω	1.3 × 7.0 × 2.7
695D106x003562T	SPRAGUE	10µF	1.3Ω	2.5 × 7.6 × 2.5
TPSC475K035R0600	AVX	4.7µF	0.6Ω	2.6 × 6.0 × 3.2

Curves are provided which show the stable ESR range as a function of load current (see Figure 7-1, Figure 7-2, Figure 7-3, and Figure 7-4).

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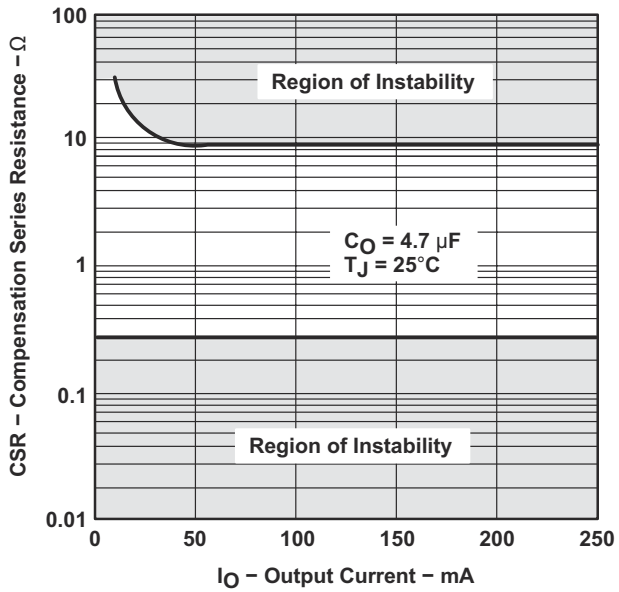


Figure 7-1. Typical Regions of Stability Compensation Series Resistance (CSR) vs Output Current (legacy chip)

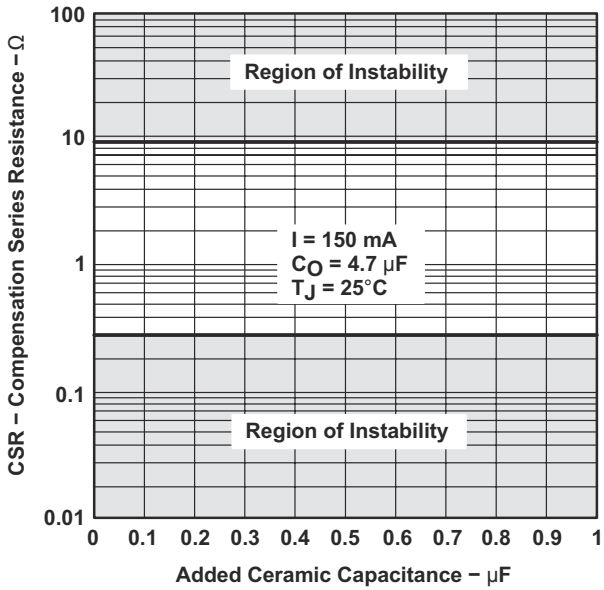


Figure 7-2. Typical Regions of Stability Compensation Series Resistance (CSR) vs Added Ceramic Capacitance (legacy chip)

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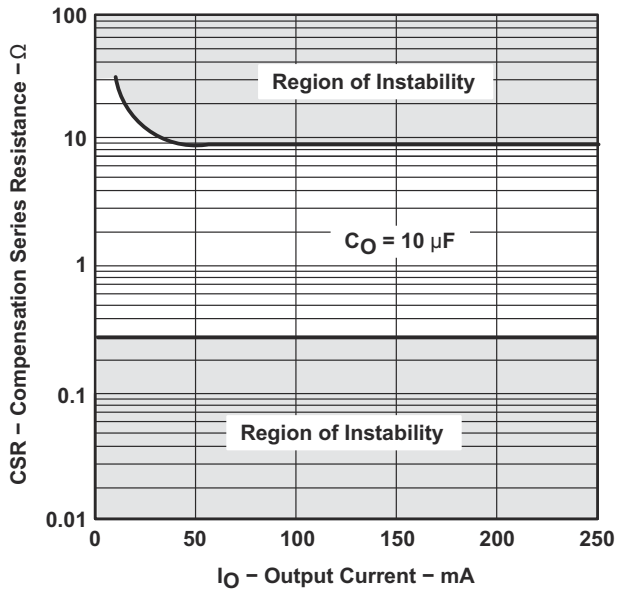


Figure 7-3. Typical Regions of Stability Compensation Series Resistance (CSR) vs Output Current (legacy chip)

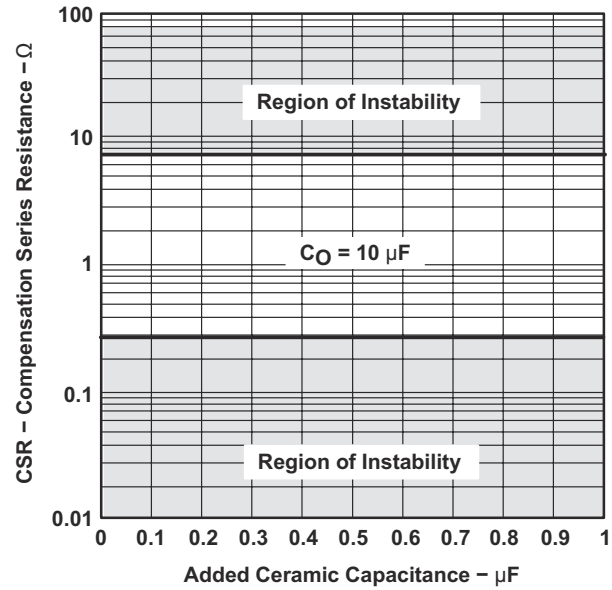


Figure 7-4. Typical Regions of Stability Compensation Series Resistance (CSR) vs Added Ceramic (legacy chip)

Remember that capacitor tolerance and variation with temperature must be taken into consideration when selecting an output capacitor so that the minimum required amount of output capacitance is provided over the full operating temperature range. The output capacitor must be located not more than 1cm from the output pin and returned to a clean analog ground.

For the new chip, dynamic performance of the device is improved with the use of an output capacitor. Use an output capacitor within the range specified in the Recommended Operating Conditions table for stability.

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7.1.4 Reverse Current

Excessive reverse current can damage this device. Reverse current flows through the intrinsic body diode of the pass transistor instead of the normal conducting channel. At high magnitudes, this current flow degrades the long-term reliability of the device.

Conditions where reverse current can occur are outlined in this section, all of which can exceed the absolute maximum rating of $V_{OUT} \leq V_{IN} + 0.3V$.

- If the device has a large C_{OUT} and the input supply collapses with little or no load current
- The output is biased when the input supply is not established
- The output is biased above the input supply

If reverse current flow is expected in the application, use external protection to protect the device. Reverse current is not limited in the device, so external limiting is required if extended reverse voltage operation is anticipated.

Figure 7-5 shows one approach for protecting the device.

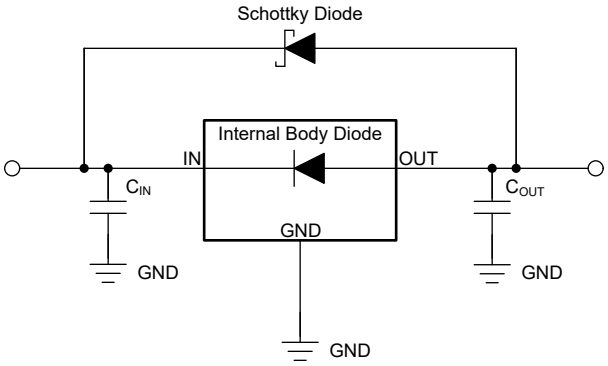


Figure 7-5. Example Circuit for Reverse Current Protection Using a Schottky Diode

7.1.5 Feed-Forward Capacitor (C_{FF})

For the adjustable-voltage version device, a feed-forward capacitor (C_{FF}) can be connected from the OUT pin to the FB pin. C_{FF} improves transient, noise, and PSRR performance, but is not required for regulator stability. Recommended C_{FF} values are listed in the *Recommended Operating Conditions* table. A higher capacitance C_{FF} can be used; however, the startup time increases. For a detailed description of C_{FF} tradeoffs, see the *Pros and Cons of Using a Feedforward Capacitor with a Low-Dropout Regulator* application report.

C_{FF} and R_1 form a zero in the loop gain at frequency f_Z, while C_{FF}, R_1, and R_2 form a pole in the loop gain at frequency f_P. C_{FF} zero and pole frequencies can be calculated from the following equations:

$$f_Z = \frac{1}{[2 \times \pi \times C_{FF} \times R_1]} \quad (3)$$

$$f_P = \frac{1}{[2 \times \pi \times C_{FF} \times (R_1 \| R_2)]} \quad (4)$$

$C_{FF} \geq 10pF$ is required for stability if the feedback divider current is less than 5µA. Equation 5 calculates the feedback divider current.

$$I_{FB_Divider} = \frac{V_{OUT}}{[R_1 + R_2]} \quad (5)$$

To avoid startup time increases from C_{FF}, limit the product $C_{FF} \times R_1 < 50\mu s$.

For an output voltage of 0.8V with the FB pin tied to the OUT pin, no C_{FF} is used.

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7.1.6 Power Dissipation (P_D)

Circuit reliability requires consideration of the device power dissipation, location of the circuit on the printed circuit board (PCB), and correct sizing of the thermal plane. The PCB area around the regulator must have few or no other heat-generating devices that cause added thermal stress.

To first-order approximation, power dissipation in the regulator depends on the input-to-output voltage difference and load conditions. The following equation calculates power dissipation (P_D).

$$P_D = [V_{IN} - V_{OUT}] \times I_{OUT} \quad (6)$$

Note

Power dissipation can be minimized, and therefore greater efficiency can be achieved, by correct selection of the system voltage rails. For the lowest power dissipation use the minimum input voltage required for correct output regulation.

For devices with a thermal pad, the primary heat conduction path for the device package is through the thermal pad to the PCB. Solder the thermal pad to a copper pad area under the device. This pad area must contain an array of plated vias that conduct heat to additional copper planes for increased heat dissipation.

The maximum power dissipation determines the maximum allowable ambient temperature (T_A) for the device. According to the following equation, power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance ($R_{\theta JA}$) of the combined PCB and device package and the temperature of the ambient air (T_A).

$$T_J = T_A + \left[R_{\theta JA} \times P_D\right] \quad (7)$$

Thermal resistance ($R_{\theta JA}$) is highly dependent on the heat-spreading capability built into the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The junction-to-ambient thermal resistance listed in the *Thermal Information* table is determined by the JEDEC standard PCB and copper-spreading area, and is used as a relative measure of package thermal performance.

7.1.7 Estimating Junction Temperature

The JEDEC standard now recommends the use of psi (Ψ) thermal metrics to estimate the junction temperatures of the linear regulator when in-circuit on a typical PCB board application. These metrics are not thermal resistance parameters and instead offer a practical and relative way to estimate junction temperature. These psi metrics are determined to be significantly independent of the copper area available for heat-spreading. The Thermal Information table lists the primary thermal metrics, which are the junction-to-top characterization parameter (ψ_{JT}) and junction-to-board characterization parameter (ψ_{JB}). These parameters provide two methods for calculating the junction temperature (T_J), as described in the following equations. Use the junction-to-top characterization parameter (ψ_{JT}) with the temperature at the center-top of device package (T_T) to calculate the junction temperature. Use the junction-to-board characterization parameter (ψ_{JB}) with the PCB surface temperature 1mm from the device package (T_B) to calculate the junction temperature.

$$T_J = T_T + \psi_{JT} \times P_D \quad (8)$$

where:

- P_D is the dissipated power
- T_T is the temperature at the center-top of the device package

$$T_J = T_B + \psi_{JB} \times P_D \quad (9)$$

where:

- T_B is the PCB surface temperature measured 1mm from the device package and centered on the package edge

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For detailed information on the thermal metrics and thermal metric use, see the *Semiconductor and IC Package Thermal Metrics* application note.

7.2 Typical Application

A typical application circuit is shown in Figure 7-6.

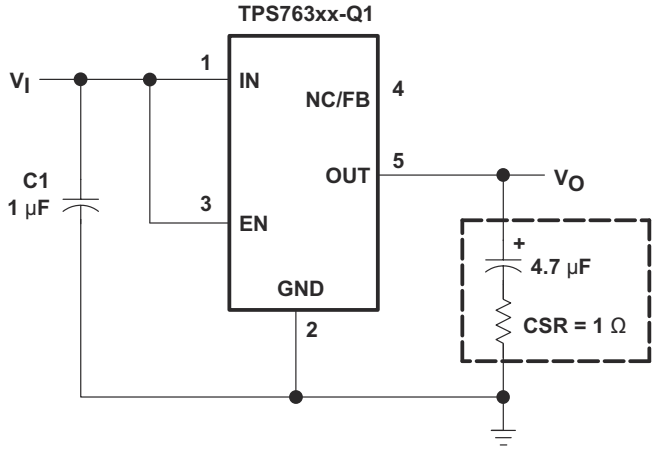


Figure 7-6. Typical Application Circuit

7.2.1 Design Requirements

Table 7-2 lists the design requirements.

Table 7-2. Design Parameters

PARAMETER	DESIGN REQUIREMENTS
Input voltage	2.7 to 10V
Output voltage	2.5 to 6.45V
Output current	0 to 150mA

7.2.2 Detailed Design Procedure

7.2.2.1 Output Voltage Programming

The output voltage of the TPS76301-Q1 adjustable regulator is programmed using an external resistor divided as shown in Figure 7-7. The output voltage is calculated using Equation 10.

$$V_O = 0.995 \times V_{REF} \times \left[1 + \frac{R1}{R2}\right] \quad (10)$$

where

- V_{REF} = 1.192V typical for legacy chip and 1.2V typical for new chip (the internal reference voltage)
- 0.995 is a constant used to center the load regulator (1%)

Choose resistors R1 and R2 for the approximately 7µA divider current. Lower value resistors can be used, but offer no inherent advantage and waste more power. Avoid higher values as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose R2 = 169kΩ to set the divider current at 7µA and then calculate R1 using Equation 11.

$$\text{Line Reg.}\,(\text{mV}) = (\%/\text{V}) \times \frac{V_O[V_{Imax} - (V_O + 1)]}{100} \times 1000 \quad (11)$$

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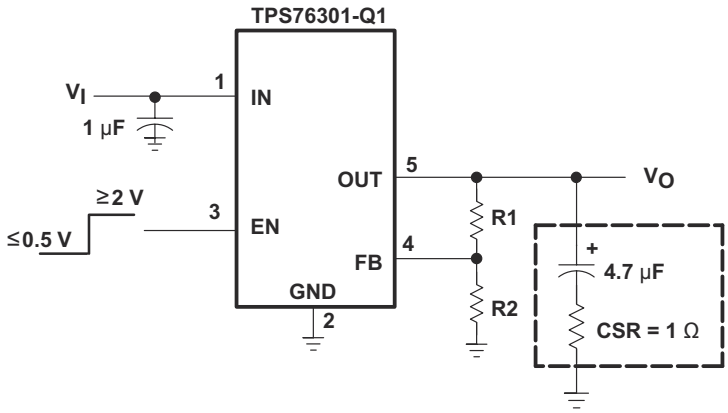


Figure 7-7. TPS76301-Q1 Adjustable LDO Regulator Programming

Table 7-3. Output Voltage Programming Guide

OUTPUT VOLTAGE (V)	DIVIDER RESISTANCE (kΩ)(1)	
	R1	R2
2.5	187	169
3.3	301	169
3.6	348	169
4	402	169
5	549	169
6.45	750	169

(1) 1% values shown.

7.2.3 Application Curves

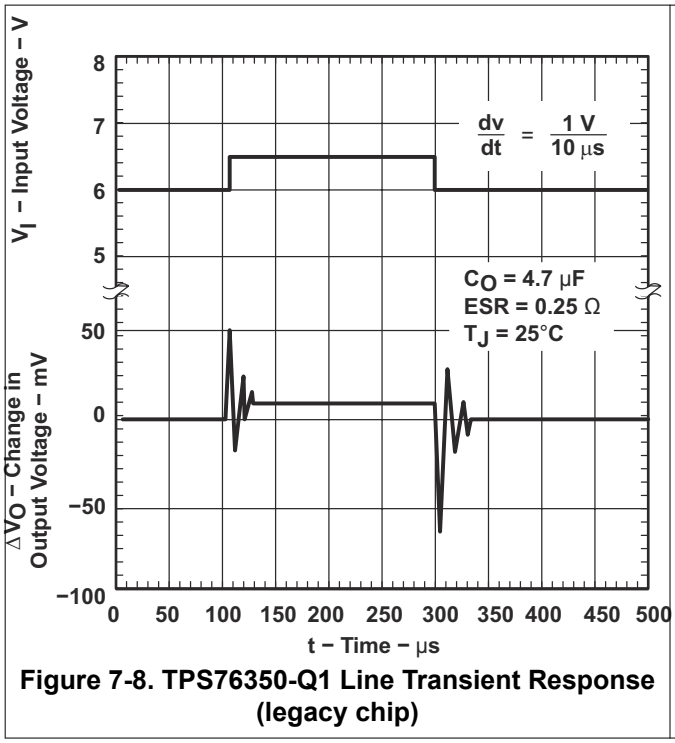


Figure 7-8. TPS76350-Q1 Line Transient Response (legacy chip)

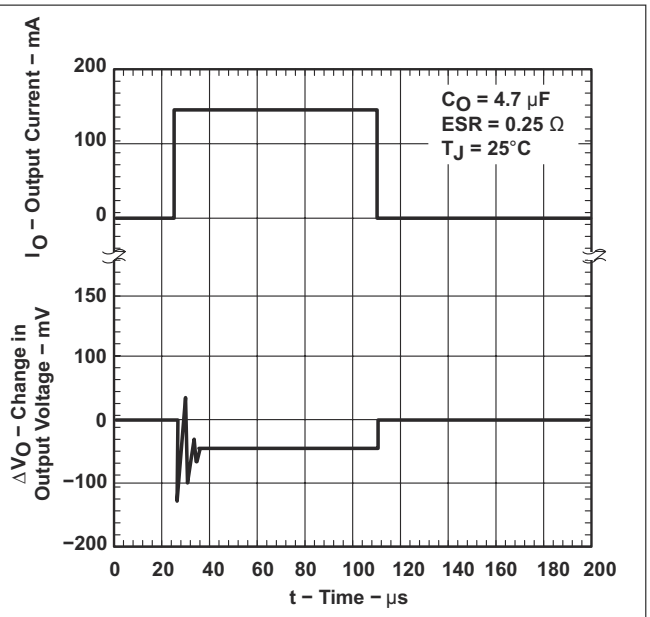
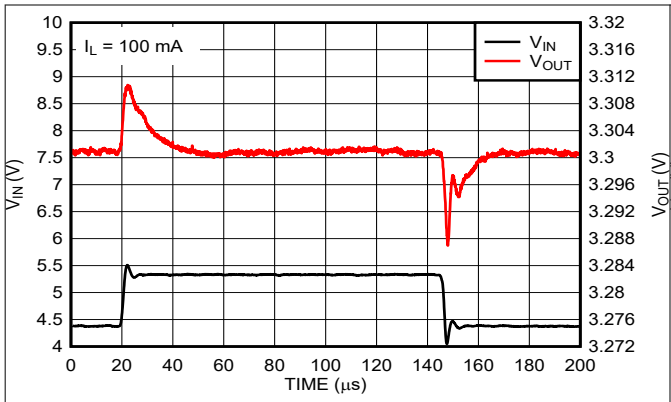


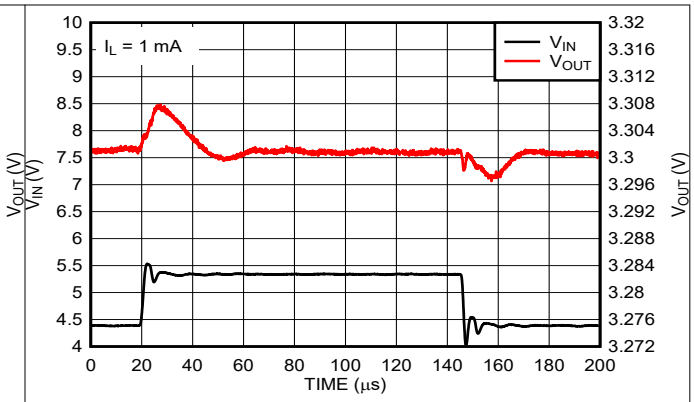
Figure 7-9. TPS76350-Q1 Load Transient Response (legacy chip)

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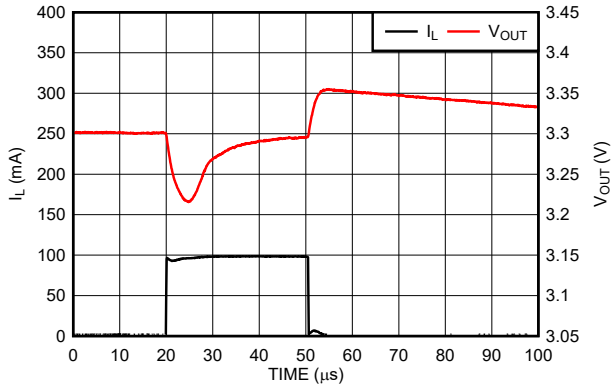
V_{OUT} = 3.3V, I_L = 100 mA

Figure 7-10. Line Transient Response (New Chip)



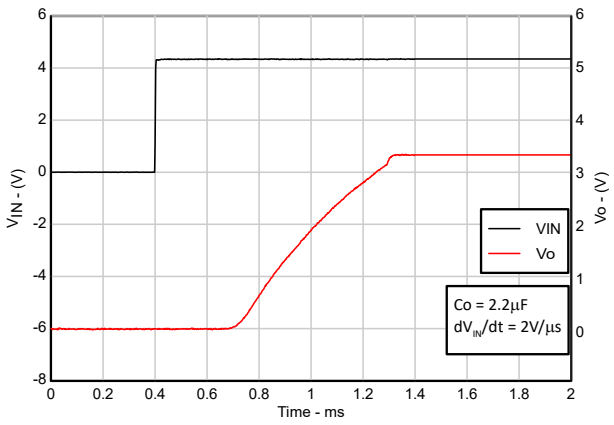
V_{OUT} = 3.3V, I_L = 1 mA

Figure 7-11. Line Transient Response (New Chip)



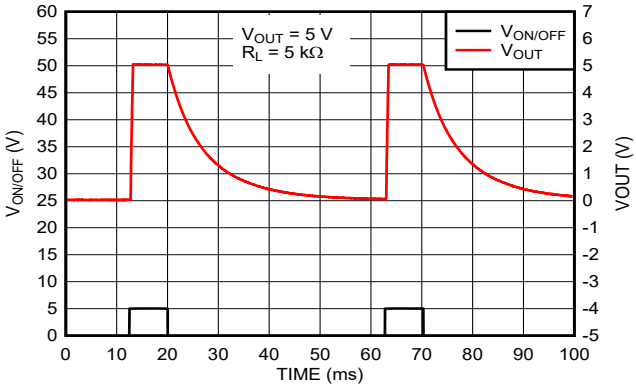
V_{OUT} = 3.3V, C_{OUT} = 2.2µF

Figure 7-12. Load Transient Response (New Chip)



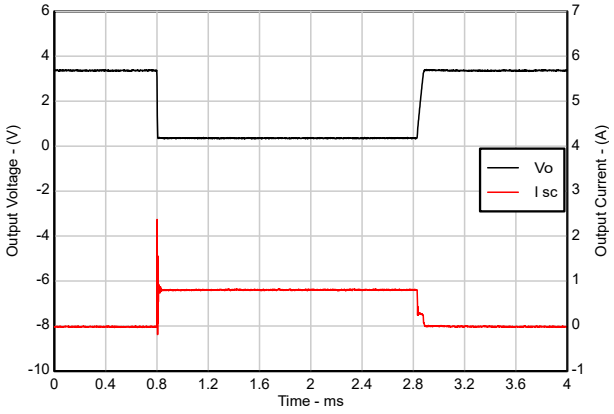
V_{OUT} = 3.3V, R_L = 3.3 kΩ

Figure 7-13. Turn-on Waveform (New Chip)



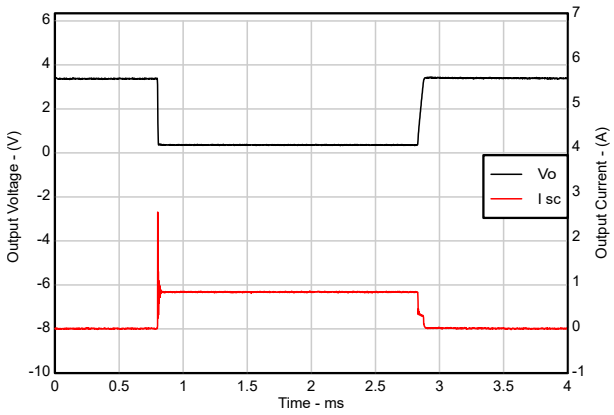
V_{OUT} = 5V, R_L = 5 kΩ

Figure 7-14. Turn-off Waveform (New Chip)



V_{IN} = 4.3V

Figure 7-15. Short Circuit Current versus Time (New Chip)



V_{IN} = 10.0V

Figure 7-16. Short Circuit Current versus Time (New Chip)

7.3 Power Supply Recommendations

These devices are designed to operate from an input voltage supply range from 2.7V to 10V. The input voltage range must provide adequate headroom in order for the device to have a regulated output. This input supply must be well-regulated and stable. Although not required, a 0.047µF or larger ceramic bypass input capacitor, connected between IN and GND and located close to the TPS763xx-Q1, is recommended to improve transient response and noise rejection. A higher-value electrolytic input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

7.4 Layout

7.4.1 Layout Guidelines

Layout is a critical part of good power-supply design. There are several signal paths that conduct fast-changing currents or voltages that interact with stray inductance or parasitic capacitance to generate noise or degrade the power-supply performance. To help eliminate these problems, bypass the IN pin to ground with a low ESR ceramic bypass capacitor with an X5R or X7R dielectric.

Minimize equivalent series inductance (ESL) and equivalent series resistance (ESR) to maximize performance and maintain stability. Place every capacitor (C_{IN}, C_{OUT}) as close as possible to the device and on the same side of the PCB as the regulator is located.

Do not place any of the capacitors on the opposite side of the PCB from where the regulator is installed. The use of vias and long traces is strongly discouraged because these circuits can impact system performance negatively, and even cause instability.

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7.4.2 Layout Example

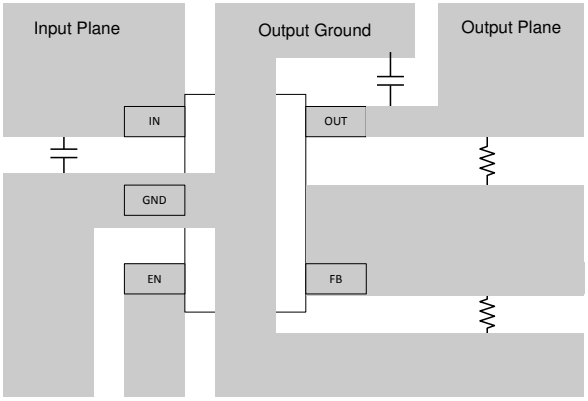


Figure 7-17. Recommended Layout

7.4.3 Power Dissipation and Junction Temperature

Specified regulator operation is designed to a junction temperature of 125°C; the maximum junction temperature allowable to avoid damaging the device is 150°C. This restriction limits the power dissipation the regulator can handle in any given application. To verify the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$ and the actual dissipation, P_D, which must be less than or equal to $P_{D(max)}$.

The maximum-power-dissipation limit is determined using Equation 12.

$$P_{D(max)} = T_{J(max)} - \frac{T_A}{R_{\theta JA}} \tag{12}$$

where

- $T_{J(max)}$ is the maximum allowable junction temperature
- $R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, see Section 5.4
- T_A is the ambient temperature

Use Equation 13 to calculate the regulator dissipation.

$$P_D = [V_I - V_O] \times I_O \tag{13}$$

Power dissipation resulting from quiescent current is negligible.

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8 Device and Documentation Support

8.1 Device Support

8.1.1 Third-Party Products Disclaimer

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8.2 Documentation Support

8.2.1 Device Nomenclature

Table 8-1. Available Options

PRODUCT(1)	V_{OUT}
TPS763**xx**Q ***yyy*** **z** *M3*Q1	**yyy** is the package designator (DBV = SOT-23-5). **y** is the reel designator size. **xx** is the nominal output voltage (for example, 3.3 = 3.3V; 5.0 = 5.0V, 01 = ADJ). This device ships with either the legacy chip (CSO: DLN or GF8) or the new chip (CSO: RFB), which uses the latest manufacturing flow. The reel packaging label provides CSO information to distinguish which chip is used. Device performance for new and legacy chips is denoted throughout the document. M3 is a suffix designator only significant for the new chip with CSO:RFB, which uses the latest manufactinno.

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

8.2.2 Related Documentation

TPS793xx-Q1 Ultralow-Noise, High-PSRR, Fast RF 200mA Low-Dropout Linear Regulators, SGLS162

8.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.4 Support Resources

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8.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.7 Glossary

TI Glossary — This glossary lists and explains terms, acronyms, and definitions.

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9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (March 2016) to Revision C (December 2025) **Page**

- Updated the numbering format for tables, figures, and cross-references throughout the document ... 1
- Added new chip information throughout the document ... 1
- Updated title from: TPS763xx-Q1 Low-Power, 150-mA, Low-Dropout Linear Regulators to: TPS763-Q1 Automotive 150mA, 10V, Low-Dropout Linear Regulator ... 1
- Updated temperature Grade 1 maximum from +125°C to +150°C in Features section ... 1
- Updated output voltage ... 1
- Deleted low-dropout regulator data ... 1
- Added device CDM ESD classification: level C4 (new chip), input voltage range, fixed and adjustable input voltage range, output current, output voltage accuracy, low quiescent current, new chip dropout voltage, supported ESR range, active over-shoot pulldown protection, new chip operating junction temperature, and °C/W packaging data ... 1
- Updated applications section ... 1
- Updated description section ... 1
- Deleted Voltage Options section ... 3
- Added nominal output capacitance note ... 3
- Updated Pin Configuration and Functions section ... 3
- Added new chip typical characteristics ... 9
- Updated overview section ... 16
- Added fixed and adjustable new chip functional block diagrams ... 16
- Deleted Regulator Protection section ... 17
- Added Output Enable section ... 17
- Added Dropout Voltage section ... 18
- Added Current Limit section ... 18
- Added Output Pulldown section ... 19
- Added Thermal Shutdown section ... 19
- Added Device Functional Mode Comparison section ... 19
- Updated Device Functional Mode Comparison table ... 19
- Moved Device Functional Mode Comparison table to Device Functional Mode Comparison section ... 19
- Updated Normal Operation section ... 20
- Added steady dropout state to Dropout Operation section ... 20
- Updated Disabled section ... 20
- Added Adjustable Device Feedback Resistors section ... 21
- Added Recommended Capacitor Types sections ... 21
- Moved External Capacitor Requirements information to Recommended Capacitor Types ... 21
- Added Reverse Current section ... 25
- Added Feed-Forward Capacitor (CFF) section ... 25
- Added Power Dissipation (PD) section ... 26
- Moved Power Dissipation and Junction Temperature information to Added Power Dissipation (PD) section ... 26
- Added Estimating Junction Temperature section ... 26
- Deleted External Capacitor Requirements section ... 27
- Updated Output Voltage Programming section ... 27
- Added new chip figures to Application Curves section ... 28

Changes from Revision * (September 2011) to Revision B (March 2016) **Page**

- Removed 3.8V, 2.8V, and 2.7V output voltage versions from the data sheet ... 1
- Removed the TPS76327-Q1, TPS76328-Q1, and TPS76338-Q1 parts from the data sheet ... 1

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- Added *ESD Ratings* table, *Feature Description* section, *Device Functional Modes*, *Application and Implementation* section, *Power Supply Recommendations* section, *Layout* section, *Device and Documentation Support* section, and *Mechanical, Packaging, and Orderable Information* section 1

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TPS76301QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAN
TPS76301QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAN
TPS76301QDBVRM3Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAN
TPS76301QDBVRQ1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAN
TPS76301QDBVRQ1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAN
TPS76316QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAO
TPS76316QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAO
TPS76318QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAP
TPS76318QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAP
TPS76318QDBVRQ1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAP
TPS76318QDBVRQ1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAP
TPS76325QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAQ
TPS76325QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAQ
TPS76330QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAT
TPS76330QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAT
TPS76333QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAU
TPS76333QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAU
TPS76333QDBVRM3Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAU
TPS76333QDBVRQ1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAU
TPS76333QDBVRQ1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAU
TPS76350QDBVRG4Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAW
TPS76350QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAW
TPS76350QDBVRM3Q1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAW
TPS76350QDBVRQ1	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAW
TPS76350QDBVRQ1.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	BAW

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF TPS763-Q1 :

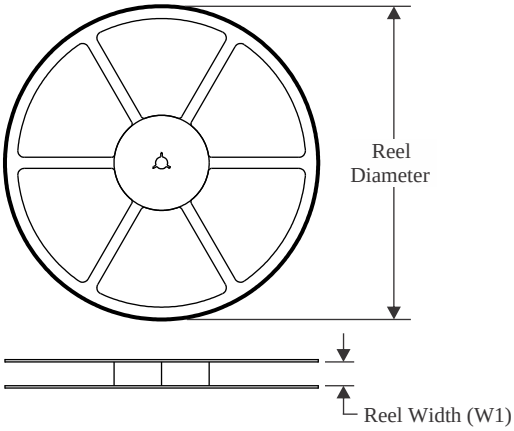
- Catalog : TPS763

NOTE: Qualified Version Definitions:

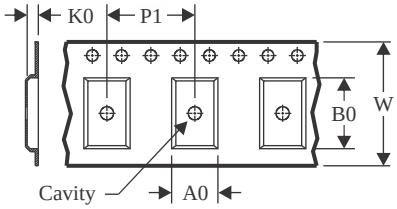
- Catalog - TI's standard catalog product

TAPE AND REEL INFORMATION

REEL DIMENSIONS

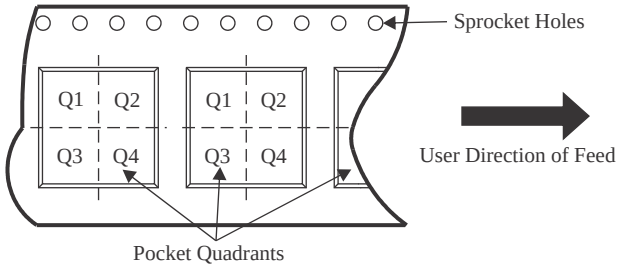


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

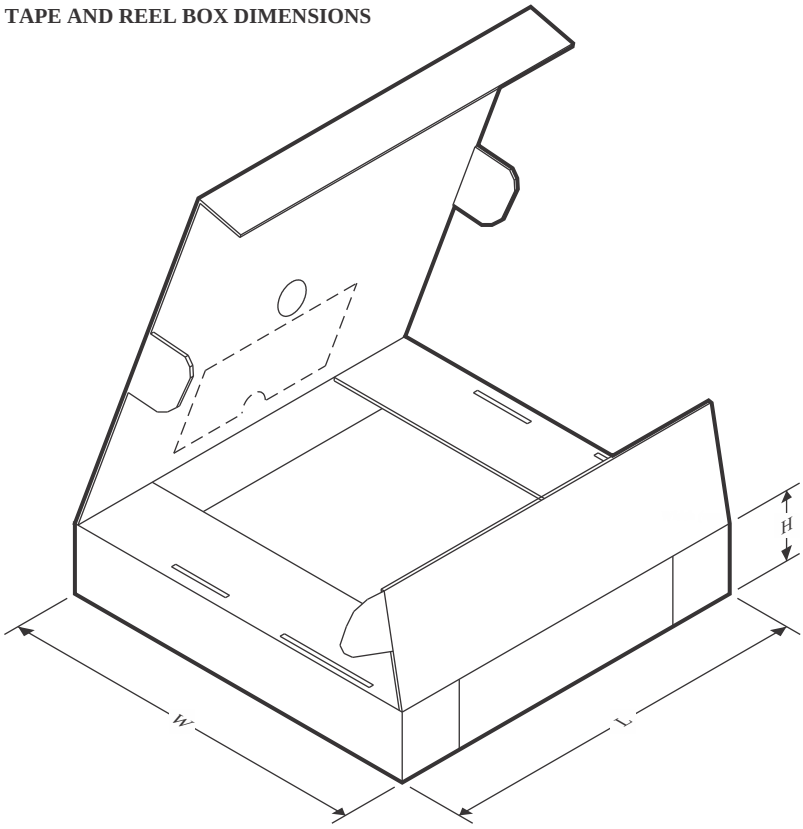


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS76301QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76301QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76301QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76301QDBVRQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76316QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76318QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76318QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76325QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76330QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76333QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76333QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76333QDBVRQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76333QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76350QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76350QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76350QDBVRQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS76350QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

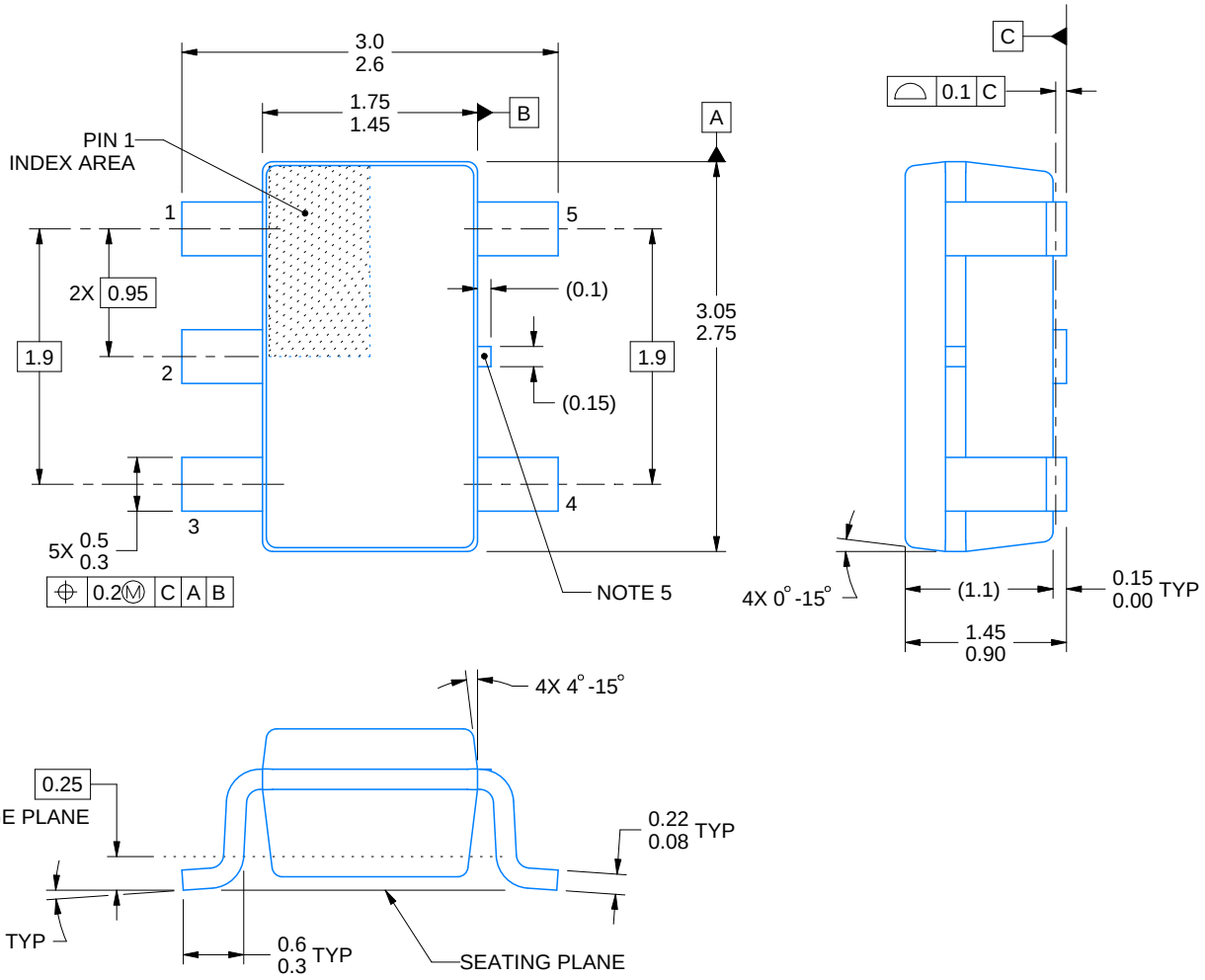
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS76301QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76301QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76301QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76301QDBVRQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76316QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76318QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76318QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76325QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76330QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76333QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76333QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76333QDBVRQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76333QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76350QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76350QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76350QDBVRQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76350QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0

PACKAGE OUTLINE

DBV0005A

SOT-23 - 1.45 mm max height

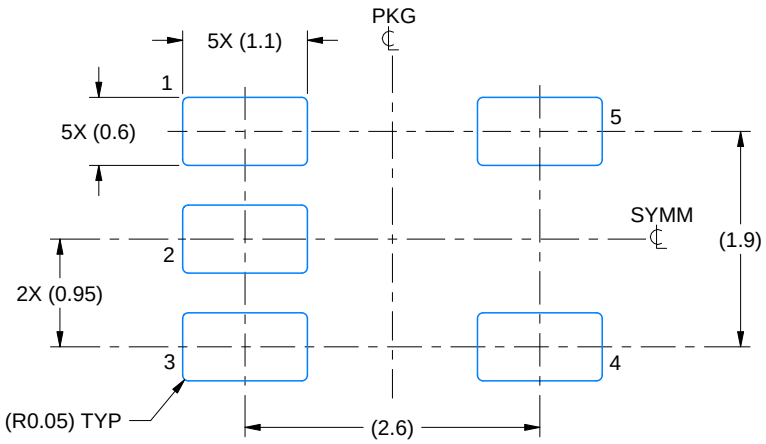
SMALL OUTLINE TRANSISTOR



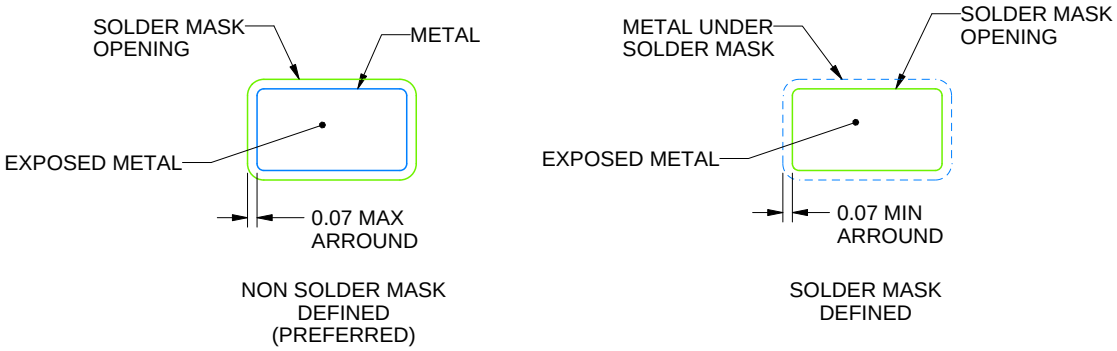
4214839/K 08/2024

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Refernce JEDEC MO-178.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X

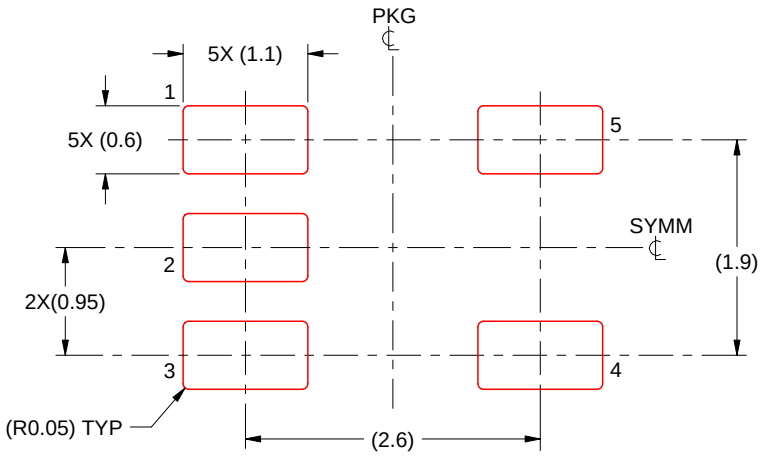


SOLDER MASK DETAILS

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NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:15X

4214839/K 08/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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Parça koduyla arama

Üretici parça numarasını yazın; stok ve fiyatı hemen görün.

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Ürün kategorileri

Entegreden kapasitöre tüm elektronik komponent kategorileri.

Kategorilere göz atın

Üreticiler

Aradığınız üreticinin ürünlerini tek sayfada inceleyin.

Üreticileri görün

Toplu alım ve teklif

Listede olmayan ya da yüksek adetli parçalar için teklif isteyin.

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